

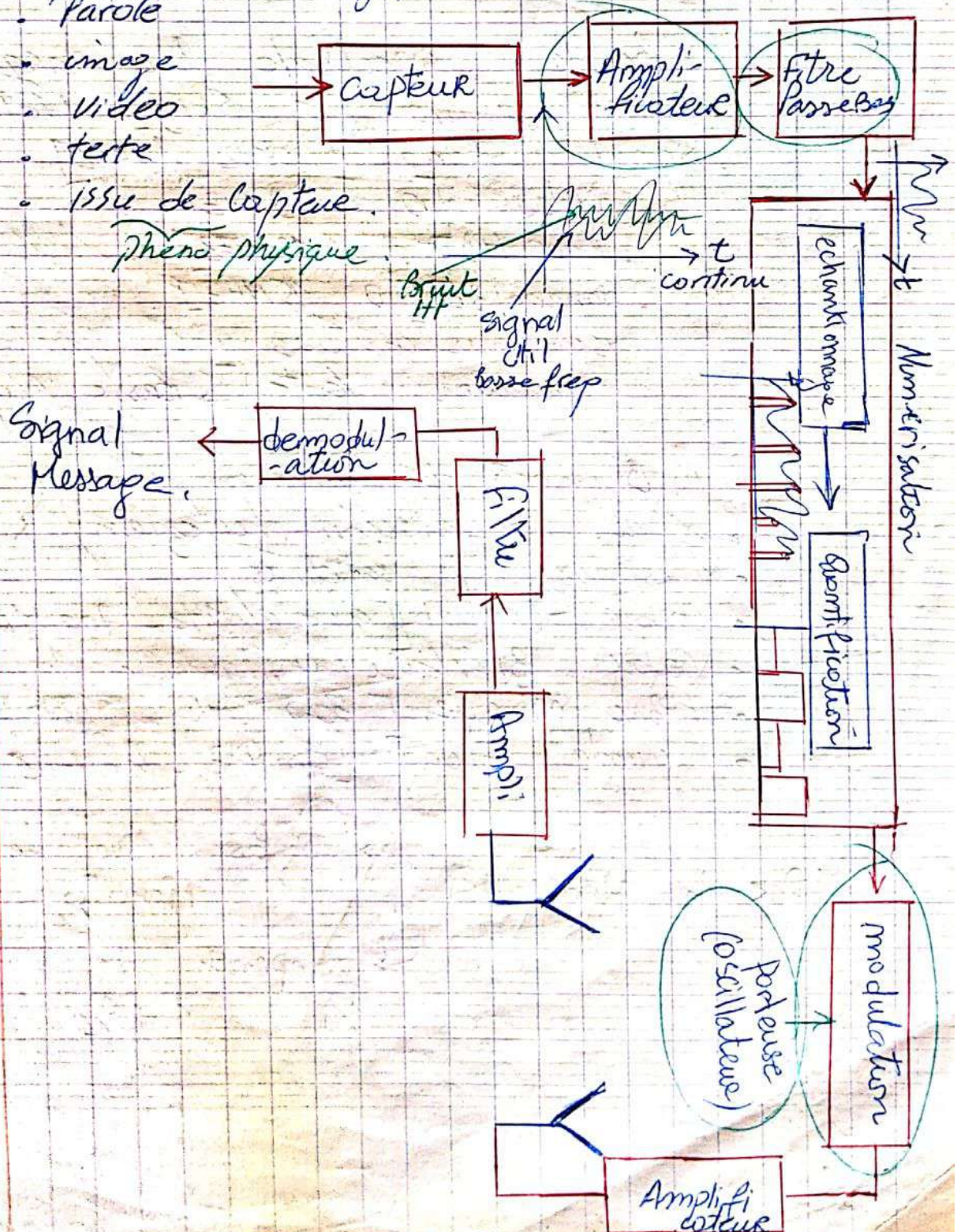
# Communication Analogique

## Chapitre 1: Introduction

### Définition

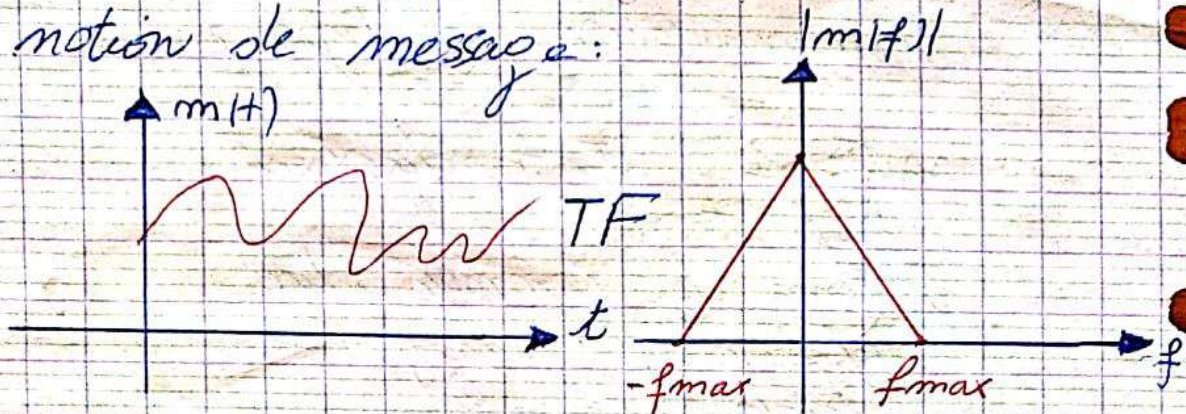
Système de Communication  
information (message)

- parole
- image
- video
- texte
- issue de capture  
phénom physique





notion de message:



Bande spectrale du msg:

- Signal audio  $[0 \text{ } 5\text{kHz}]$  (téléphonie)
- Signal stéréo  $[20\text{kHz} \text{ } 20\text{kHz}]$
- Vidéo  $[0 \text{ } 5\text{MHz}]$
- Vidéo HD  $[0 \text{ } 10\text{MHz}]$
- image 3D  $[0 \text{ } 15\text{MHz}]$
- issus des capteurs  $[0 \text{ } \text{quelques Hz}]$   
(variation lente)  $\rightarrow$  basse freq

Supports de transmission:

- d'air (faisceau hertzien)
- l'espace (satellites)
- câble coaxial  $[quelques \text{MHz} \text{ } 700 \text{MHz}]$
- câble bifilaire  $[0 \text{ } \text{quelques } 100 \text{en kHz}]$
- fibre optique  $[300 \text{GHz} \text{ } 1500 \text{GHz}]$
- guide d'ondes  $[700 \text{MHz} \text{ } 300 \text{GHz}]$

Prk la Modulation  $\rightarrow$  ne pas perdre le msg (Parasites ...)

$\rightarrow$  depends de l'antenne  
 $\lambda/4$   $\lambda = c/f$



## Modulation Analogique.

AM, FM, PM (Radio)

- AM, DSB (porteur supprimée) → Mixeur.
- BLU Bande Latérale unique → navigation
- BLI Bande Latérale indépendante, stéréo
- BLA " " (Atténuée) → télévision
- QAM (Quadrature) Codage Audio Visuel.

## Modulation Numérique:

PSK, ASK, FSK, QPSK

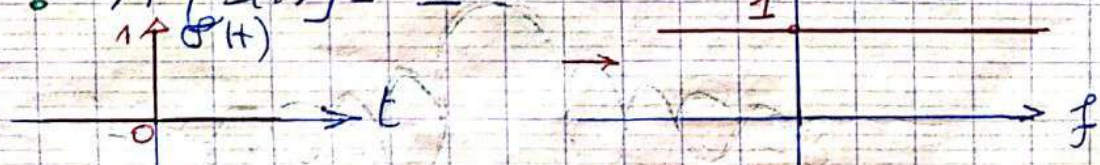
19-2019

### Spéctre

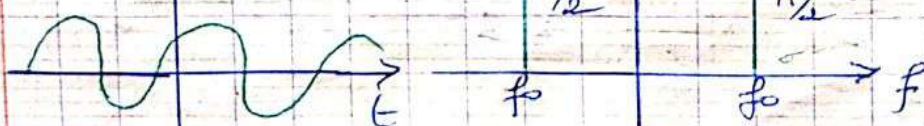
L'ensemble des fréquences auxquelles se compose le signal.

### propriétés:

- $x(t) \xrightarrow{TF} X(f)$
- $x(t - t_0) \rightarrow X(f) e^{-j2\pi f t_0}$
- $TF[x_1(t) + x_2(t)] = TF[x_1(t)] + TF[x_2(t)]$   
 $= X_1(f) + X_2(f)$
- $TF[\delta(t)] = 1$



$$TF[A \cos 2\pi f_0 t] = \frac{A}{2} (\delta(f - f_0) + \delta(f + f_0))$$



$$TF\left[\frac{A}{2} \sin 2\pi f_0 t\right] = \frac{A}{2j} (\delta(f - f_0) - \delta(f + f_0))$$



Exemple:



$$X(f) = \int_{-\infty}^{+\infty} A \operatorname{rect}\left(\frac{t}{T}\right) e^{-j2\pi f t} dt$$

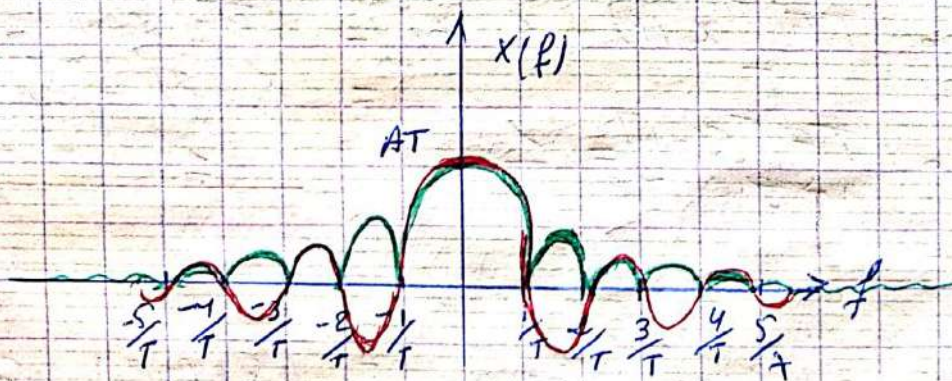
$$= \int_{-T/2}^{T/2} A e^{-j2\pi f t} dt$$

$$= \frac{A}{-j2\pi f} \left[ e^{-j2\pi f t} \right]_{-T/2}^{T/2} = \frac{A}{-j2\pi f} \left( e^{-j\pi f T} - e^{j\pi f T} \right)$$

$$X(f) = \frac{A}{-j2\pi f} \left( \cos \pi f T - j \sin \pi f T - \cos \pi f T - \sin \pi f T \right)$$

$$X(f) = \frac{A}{-j2\pi f} (-j2 \sin \pi f T) = \frac{AT}{\pi f T} \sin \pi f T$$

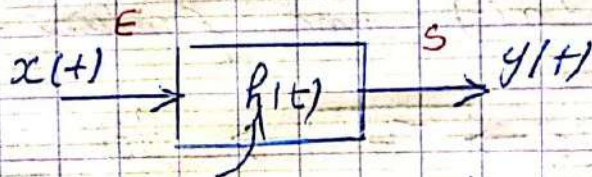
$$= AT \frac{\sin \pi f T}{\pi f T} = \boxed{AT \operatorname{sinc} fT}$$



$$\lim_{x \rightarrow 0} \frac{\sin x}{x} = 1$$



# Convolution



Réponse impulsionnelle.

$H(f)$  = Fonction de transfert.

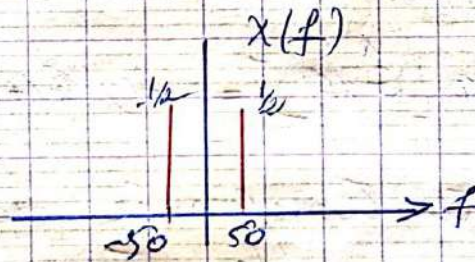
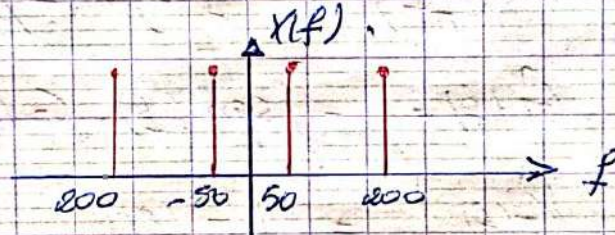
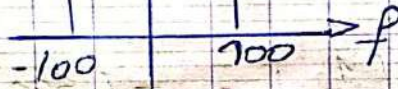
$$H(f) = S/E$$

Exemple de filtrage:

$$x(t) = \cos 100 \pi t + \cos 400 \pi t$$

$R(t)$

TF



$$y(t) = TF^{-1} \{ Y(f) \} = TF^{-1} \left[ \frac{1}{2} \delta(f+50) + \frac{1}{2} \delta(f-50) \right] = \cos 100 \pi t$$



Transformée de Hilbert :  
Soit  $g(t)$  une fonction :

$$g(t) \rightarrow \boxed{TH} \rightarrow \hat{g}(t)$$

$$\hat{g}(t) = g(t) * \frac{1}{\pi t} = \int_{-\infty}^{+\infty} g(\tau) \frac{1}{\pi(t-\tau)} d\tau$$

propriétés :

$$\begin{aligned} g(t) &\rightarrow G(f) \\ \hat{g}(t) &= \hat{G}(f) \\ \hat{G}(f) &= G(f) \cdot TF\left(\frac{1}{\pi f}\right) \end{aligned}$$

$$TF\left(\frac{1}{\pi f}\right) = -j \operatorname{sgn}(f)$$

$$\hat{G}(f) = -j \operatorname{sgn}(f) \cdot G(f)$$

$$|\hat{G}(f)| = |G(f)|$$

$$\operatorname{Arg}[\hat{G}(f)] = \operatorname{Arg}(G(f)) + \pi/2$$

Cette transformée déphase le signal  
de  $-\pi/2$



transformée de Hilbert :

TF, TL, 3  $t \rightarrow f$ TH  $t \rightarrow t$ 

$$\hat{g}(t) = \text{TH} [g(t)] = g(t) * \frac{1}{\pi t}$$

$$\hat{g}(t) = \int_{-\infty}^{+\infty} g(\tau) \cdot \frac{1}{\pi(t-\tau)} d\tau$$

spectre de  $\hat{g}(t)$  :

$$\begin{aligned} \hat{G}(f) &= \text{TF} [\hat{g}(t)] \\ &= \text{TF} \left[ g(t) * \frac{1}{\pi t} \right] \end{aligned}$$

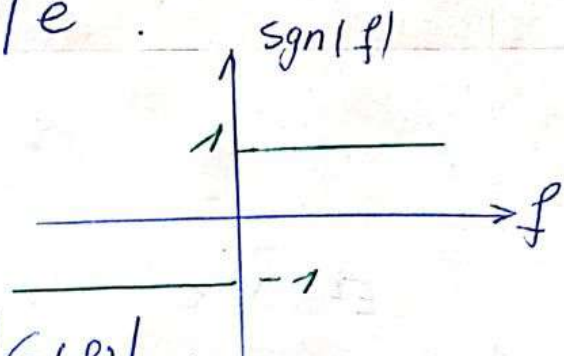
$$\hat{G}(f) = G(f) \cdot \text{TF} \left[ \frac{1}{\pi t} \right]$$

$$\text{OR : } \text{TF} \left[ \frac{1}{\pi t} \right] = -j \text{sgn}(f)$$

$$\hat{G}(f) = -j \text{sgn}(f) \cdot G(f)$$

$$\hat{G}(f) = |\hat{G}(f)| e^{-j\pi/2}$$

$$|\hat{G}(f)| = ?$$



$$\Rightarrow |\hat{G}(f)| = |G(f)| = 1$$

L'argument

$$\begin{aligned} \text{Arg}(\hat{G}(f)) &= -\frac{\pi}{2} + 0 + \text{Arg}(G(f)) \\ &= \text{Arg}(G(f)) - \frac{\pi}{2} \end{aligned}$$

elle garde le signal (spectre Mod Amp) mais le fait retarder de  $\frac{\pi}{2}$



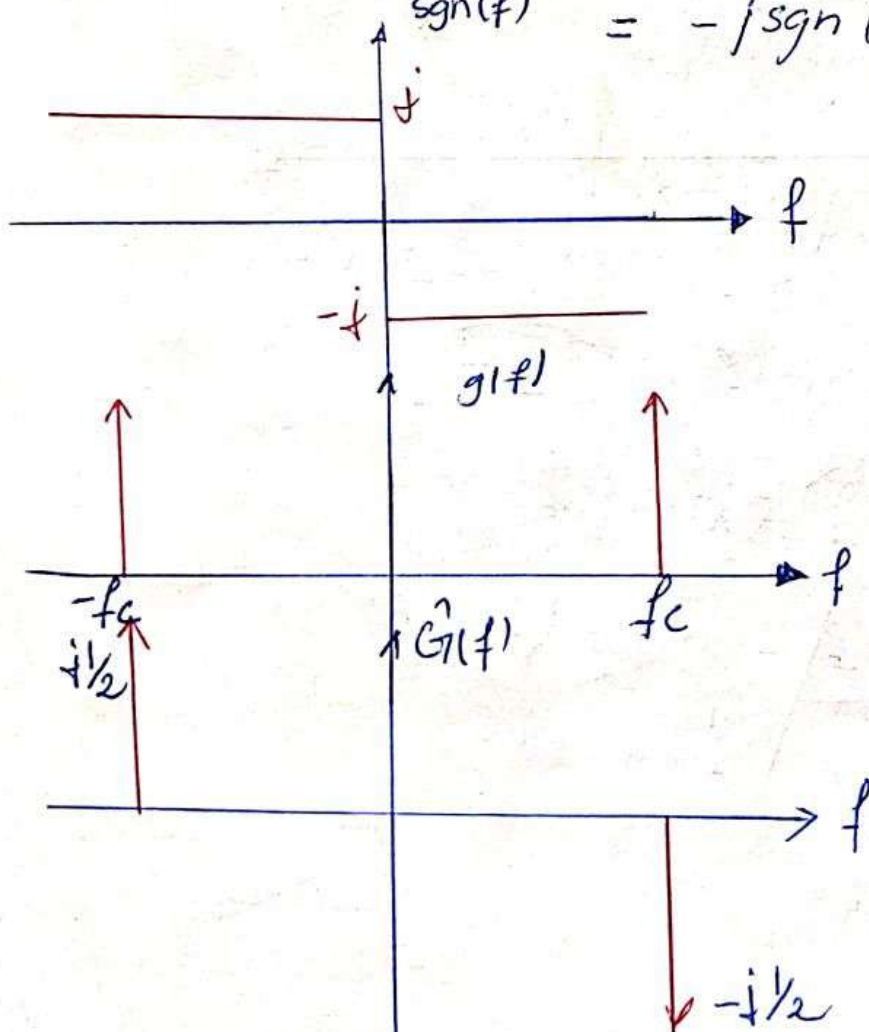
Exemple:  $g(t) = \cos 2\pi f_c t$ .

$$\begin{aligned} TH[\cos 2\pi f_c t] &= \cos 2\pi f_c t * \frac{1}{\pi t} \\ &= \int_{-\infty}^{+\infty} (\cos 2\pi f_c \tau) * \frac{1}{\pi(t-\tau)} d\tau. \end{aligned}$$

$$\hat{G}(f) \xrightarrow{TF} \hat{g}(t).$$

$$\hat{G}(f) = TF(\cos 2\pi f_c t) \cdot TF\left(\frac{1}{\pi t}\right).$$

$$\text{sgn}(f) = -j \text{sgn}(f) \cdot \left( \frac{1}{2} (\delta(f-f_c) + \delta(f+f_c)) \right)$$



impléxe.

Et Donc :

$$\begin{aligned} \hat{G}(f) &= \left( -j \frac{1}{2} (\delta(f-f_c)) + j \frac{1}{2} (\delta(f+f_c)) \right) \\ &= \frac{1}{j2} (\delta(f-f_c) - \delta(f+f_c)) \\ &= \frac{1}{j2} (\delta(f-f_c) - \delta(f+f_c)) \end{aligned}$$



$$= TF (\sin \pi f c t).$$

Forme Canonique d'un Signal :

notion d'un Signal Analytique :

$$g_+(t) = g(t) + j \hat{g}(t).$$

$|g_+(t)|$  : pré enveloppe.

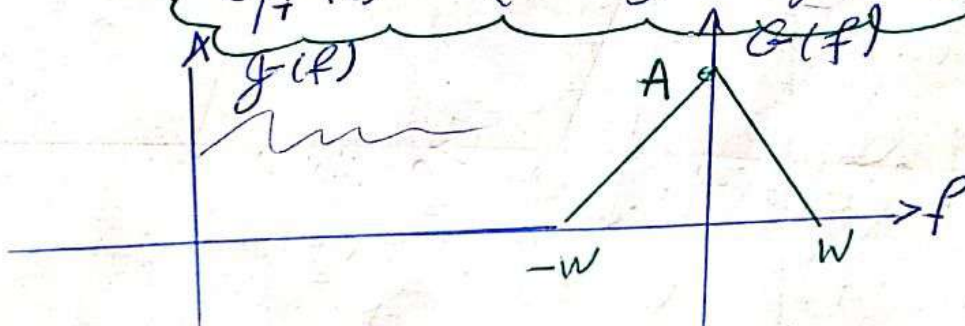
$$G_+(f) \xleftrightarrow{TF} g_+(t).$$

$$G_+(f) = G(f) + j \hat{G}(f).$$

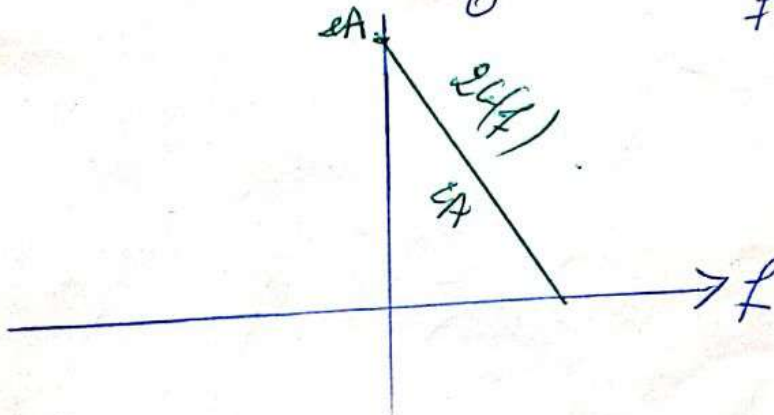
$$= G(f) + j (-j \operatorname{sgn}(f) G(f))$$

$$= G(f) (1 + \operatorname{sgn}(f))$$

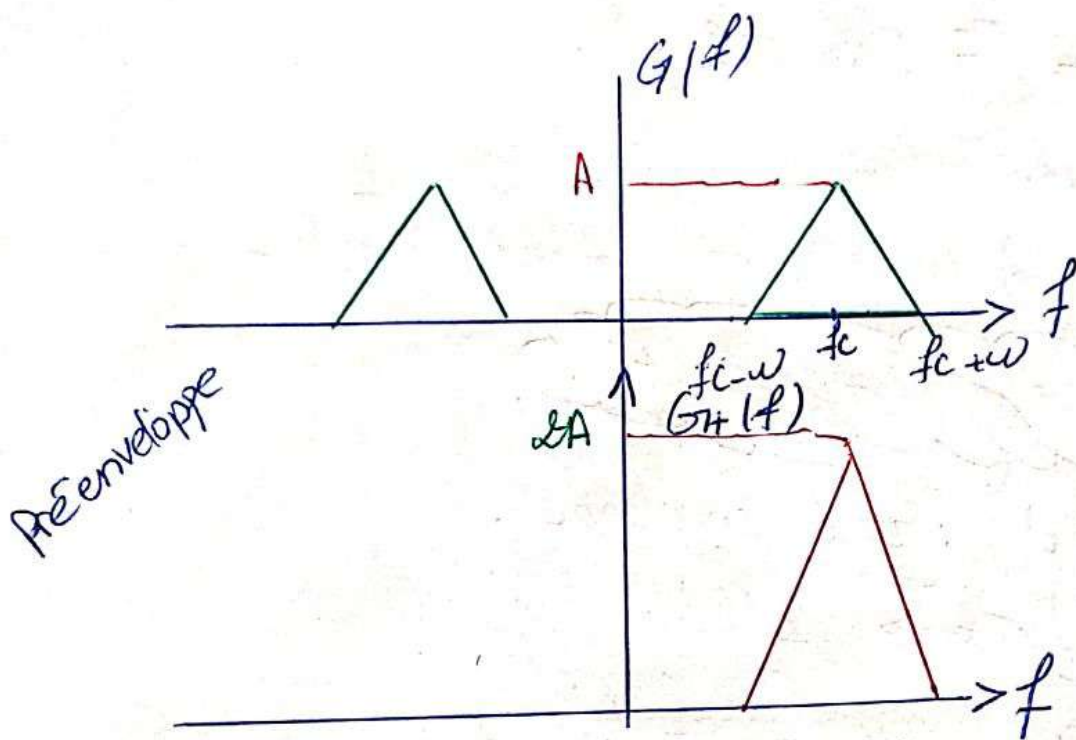
$$G_+(f) = (1 + \operatorname{sgn}(f)) \cdot G(f).$$



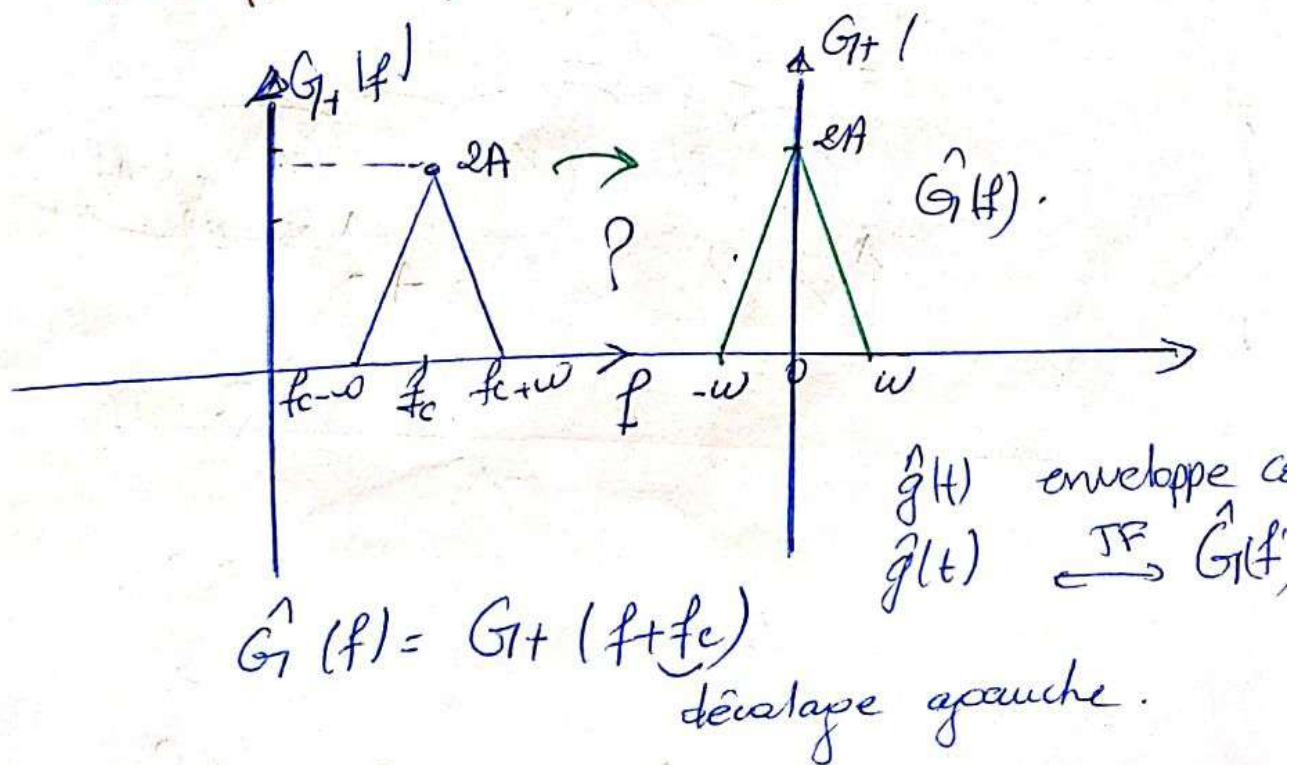
$$G_+ f = \begin{cases} 2G(f) & f > 0 \\ G(0) & f = 0 \\ 0 & f < 0 \end{cases}$$







enveloppe complexe:



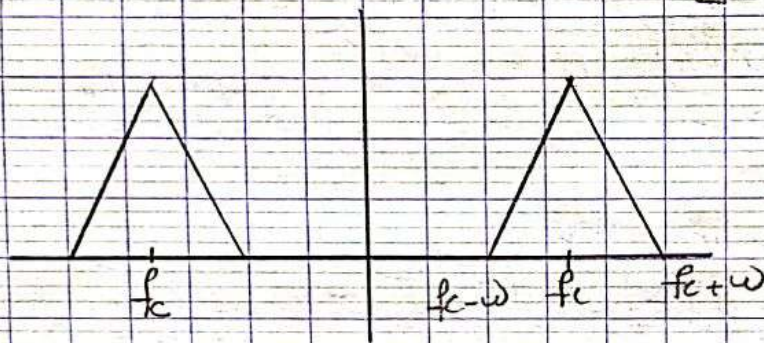


pré-enveloppe:

$$\hat{g}_+(t) = g(t) + j\hat{g}(t)$$

$$\hat{g}(t) = \mathcal{H}[g(t)]$$

$\hat{g}(t)$  c'est  $g(t)$  déphasé de  $-\frac{\pi}{2}$

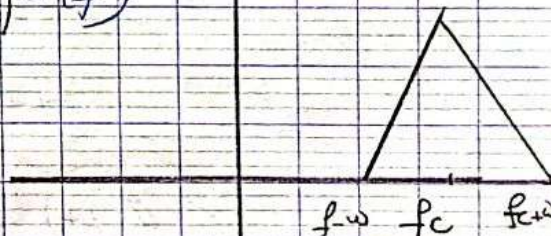


OK  $G_+(f) = G(f) + jG^\perp(f)$

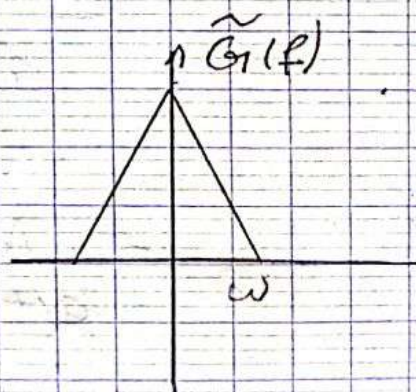
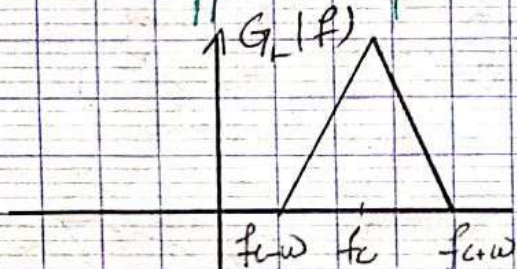
$$G^\perp(f) = -j \operatorname{sgn}(f) G(f)$$

$$G(f) = \frac{1 + \operatorname{sgn}(f)}{2} G(f)$$

$$G(f) = \begin{cases} 2b(f) & f > 0 \\ G(0) & f = 0 \\ 0 & f < 0 \end{cases}$$



Enveloppe Complexe



$$\tilde{G}(f) = G_+(f + f_c)$$

enveloppe Complexe



$$\hat{g}(t) = \text{TF}^{-1} [\hat{G}(t)] = \text{TF}^{-1} [G_+(f + f_c)]$$

On sait que :

$$\begin{aligned} x(t) &\rightarrow X(f) \\ X(f) &\rightarrow x(t) \\ X(f - f_0) &\rightarrow x(t) e^{j2\pi f_0 t} \\ x(t) &\rightarrow X(f) \\ x(t - t_0) &\rightarrow X(f) e^{-j2\pi f t_0} \end{aligned}$$

Donc  $\text{TF}^{-1}$

$$\begin{aligned} \tilde{g}(t) &= \text{TF}^{-1} [G_+(f + f_c)] \\ g(t) &= \tilde{g}(t) \cdot e^{-j2\pi f_c t} \end{aligned}$$

enveloppe complexe.

On a un signal

$$\begin{aligned} g(t) &= g(t) + j \tilde{g}(t) \\ g(t) &= \text{Signal} \\ g(t) &= \text{Re} [g_+(t)] \end{aligned}$$

$$g_+(t) = \tilde{g}(t) e^{j2\pi f_c t}$$

$\tilde{g}(t)$  est une fct complexe

$$\tilde{g}(t) = \text{Re} \tilde{g} + j \text{Im} \tilde{g}$$

et donc

$$\begin{aligned} g(t) &= \text{Re} [( \text{Re} \tilde{g} + j \text{Im} \tilde{g} ) e^{j2\pi f_c t}] \\ &= \text{Re} [ ( \text{Re} \tilde{g} + j \text{Im} \tilde{g} ) (\cos 2\pi f_c t + j \sin 2\pi f_c t) ] \\ &= \text{Re} [ \text{Re} \tilde{g} \cos 2\pi f_c t - \text{Im} \tilde{g} \sin 2\pi f_c t + j \text{Im} \tilde{g} \cos 2\pi f_c t + j \text{Re} \tilde{g} \sin 2\pi f_c t ] \end{aligned}$$



$$g(t) = \text{Re} \left[ \text{Re} \tilde{g} \cos 2\pi f_c t + \text{Im} \tilde{g} \sin 2\pi f_c t + j [\text{Re} \tilde{g} \sin 2\pi f_c t + \text{Im} \tilde{g} \cos 2\pi f_c t] \right]$$

$$g(t) = \text{Re} \tilde{g} \cos 2\pi f_c t - \text{Im} \tilde{g} \sin 2\pi f_c t$$

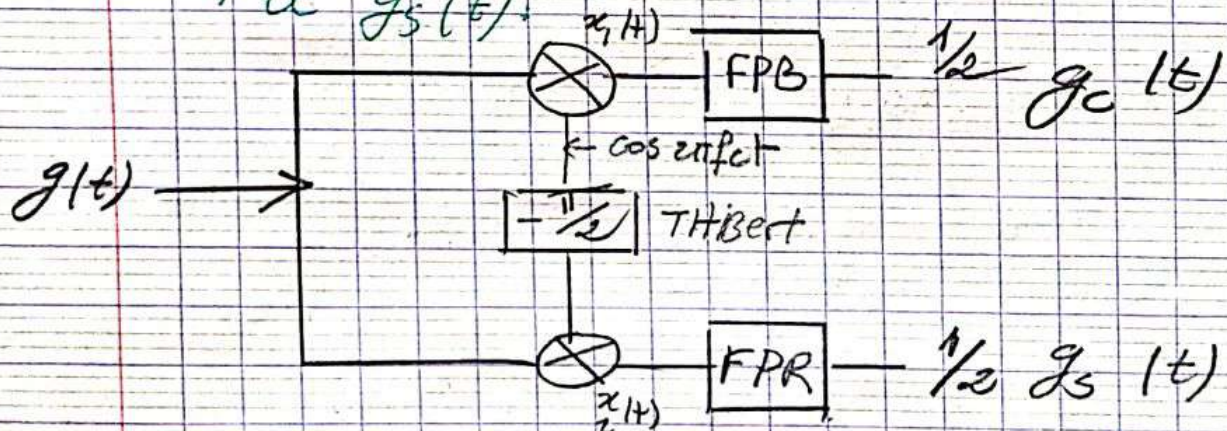
$$g(t) = g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t$$

forme canonique d'un signal modulé

$g_c(t)$  : Composante en phase par rapport à la porteuse

$g_s(t)$  : Composante en quadrature de phase

Génération de Composante  $g_c(t)$  et  $g_s(t)$



$$g(t) = g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t$$

$$x_1(t) = (g(t) \cdot \cos 2\pi f_c t) = (g_c(t) \cos^2 2\pi f_c t - g_s(t) \sin 2\pi f_c t \cdot \cos 2\pi f_c t)$$

$$\cos^2 x = \frac{1 + \cos 2x}{2} ; = \frac{g_c(t)}{2} (1 + \cos 4\pi f_c t) - \frac{g_s(t) \sin 4\pi f_c t}{2}$$



$$\frac{g_c(t)}{2} + \frac{g_c(t)}{2} \cos 4\pi f_c t - \frac{g_s}{2} \sin 4\pi f_c t$$

$$x(t) = g(t) \sin 2\pi f_c t = \frac{g_c(t)}{\sin 2\pi f_c t} \cos 2\pi f_c t \cdot \sin 2\pi f_c t$$

$$= \frac{g_c(t)}{2} \sin 4\pi f_c t - \frac{g_s(t)}{2} (1 - \cos 4\pi f_c t)$$

$$\cos^2 x = \frac{1 + \cos 2x}{2} = \frac{-g_s(t)}{2} + \frac{g_c(t)}{2} \sin 4\pi f_c t + \frac{g_s(t)}{2} \cos 4\pi f_c t$$

$$\sin^2 x = \frac{1 - \cos 2x}{2}$$



forme canonique  $g(t)$ :

$$g(t) = g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t.$$

$g(t)$  signal de hautes fréquences.

$g_c(t); g_s(t)$  " Basses fréquences

forme rectangulaire:

enveloppe complexe:

$$\tilde{g}(t) = g_+(t) e^{-j2\pi f_c t}.$$

$\tilde{g}(t) \rightarrow g(t)$  forme rectangulaire.

$$\tilde{g}(t) = g_c(t) + j g_s(t). \quad \text{On sait que: } g(t) = \text{Re} [\tilde{g}(t)]$$

$$\tilde{g}(t) = g_+(t) e^{-j2\pi f_c t} \quad \text{avec } g_+(t) = g(t) + j \tilde{g}(t).$$

$$\tilde{g}(t) = [g(t) + j \tilde{g}(t)] e^{-j2\pi f_c t}$$

or:

$$g(t) = g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t.$$

$$\begin{aligned} \tilde{g}(t) &= (g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t) + j \\ &\quad \text{TH} [g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t] e^{-j2\pi f_c t} \\ &= [g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t + \\ &\quad j \text{TH} [g_c(t) \cos 2\pi f_c t] - j \text{TH} [g_s(t) \sin 2\pi f_c t]] e^{-j2\pi f_c t} \end{aligned}$$

$$\text{TH} [m(t) \cdot c(t)]$$

$$\text{TH} = m(t) \cdot \hat{c}(t)$$

(1)5



$$TH [\cos 2\pi f_c t] = \sin 2\pi f_c t$$

$$TH [\sin 2\pi f_c t] = -\cos 2\pi f_c t$$

$$TH [g_c(t) \cos 2\pi f_c t] = g_c(t) \sin 2\pi f_c t$$

$$TH [g_s(t) \sin 2\pi f_c t] = -g_s(t) \cos 2\pi f_c t$$

$$\Rightarrow \tilde{g}(t) = (g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t + j g_c(t) \sin 2\pi f_c t + j g_s(t) \cos 2\pi f_c t) \cdot (\cos 2\pi f_c t - j \sin 2\pi f_c t)$$

$$\tilde{g}(t) = g_c(t) \cos^2 2\pi f_c t + j g_c(t) \cos 2\pi f_c t \sin 2\pi f_c t - g_s(t) \sin 2\pi f_c t \cos 2\pi f_c t + j g_s(t) \sin^2 2\pi f_c t + j g_c(t) \cos 2\pi f_c t \sin 2\pi f_c t + g_c(t) \sin^2 2\pi f_c t + j g_s(t) \cos^2 2\pi f_c t + g_s(t) \cos 2\pi f_c t \sin 2\pi f_c t$$

$$\tilde{g}(t) = g_c(t) + j g_s(t)$$

Forme triangulaire

$$\tilde{g}(t) = \sqrt{g_c(t)^2 + g_s(t)^2} \cdot e^{j\mathcal{E}}$$

$$\mathcal{E} = \arctg \frac{g_s}{g_c}$$

$$a(t) = \sqrt{g_c^2 + g_s^2}$$



$$\tilde{g}(t) = a(t) e^{j\epsilon}$$

$$g(t) = \text{Re}(g_+(t)) \text{ or } g_+(t) = g(t) + j\hat{g}(t)$$

$$g_+(t) = \tilde{g}(t) \cdot e^{+j2\pi f_c t} \quad \tilde{g}(t) g_+(t) e^{-j2\pi f_c t}$$

$$g(t) = \text{Re}[\tilde{g}(t) e^{j2\pi f_c t}] = \text{Re}[a(t) e^{j(2\pi f_c t + \epsilon)}]$$

$$g(t) = \text{Re}[a(t) e^{j(2\pi f_c t + \epsilon)}]$$

$$= \text{Re}[a(t) \cos 2\pi f_c t + j a(t) \sin 2\pi f_c t]$$

$$g(t) = a(t) \cos(2\pi f_c t + \epsilon)$$

enveloppe  
d'amplitude.

Chapitre 02: Modulation d'Amplitude

$m(t)$  message  $[-W, W]$  Avec Porteuse.  
basse fréquence.

$$c(t) = A_c \cos 2\pi f_c t$$

$$g(t) = \text{signal modulé}$$

$$g(t) = A_c (1 + k m(t)) \cos 2\pi f_c t$$

$k$ : sensibilité du modulateur.

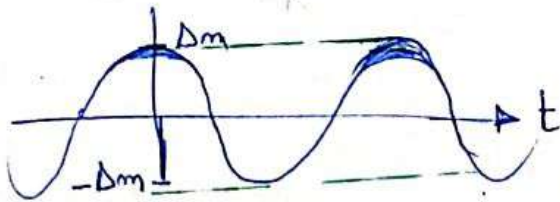
$\pm$  identifier  $g_c(t)$  et  $g_s(t)$ .

$$g_c(t) = A_c (1 + k m(t))$$

$$g_s(t) = 0$$



$$\cos a \cos b = \frac{1}{2} [\cos(a+b) + \cos(a-b)]$$



$$g(t) = A_c (1 + K A_m \cos 2\pi f_m t) \cos 2\pi f_c t$$

$$g(t) = A_c \cos 2\pi f_c t + A_c m \cos 2\pi f_c t \cdot \cos 2\pi f_m t$$

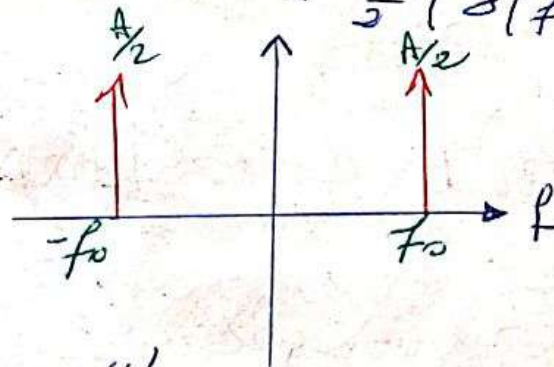
SPC), cos (y)

$$g(t) = A_c \cos 2\pi f_c t + A_c m \cos 2\pi f_c t \cdot \cos 2\pi f_m t$$

$$= A_c \cos 2\pi f_c t + \frac{A_c m}{2} \cos 2\pi (f_c - f_m) t + \frac{A_c m}{2} \cos 2\pi (f_c + f_m) t$$

On sait que:

$$TF[A \cos(2\pi f_0 t)] = \frac{A}{2} (\delta(f - f_0) + \delta(f + f_0))$$

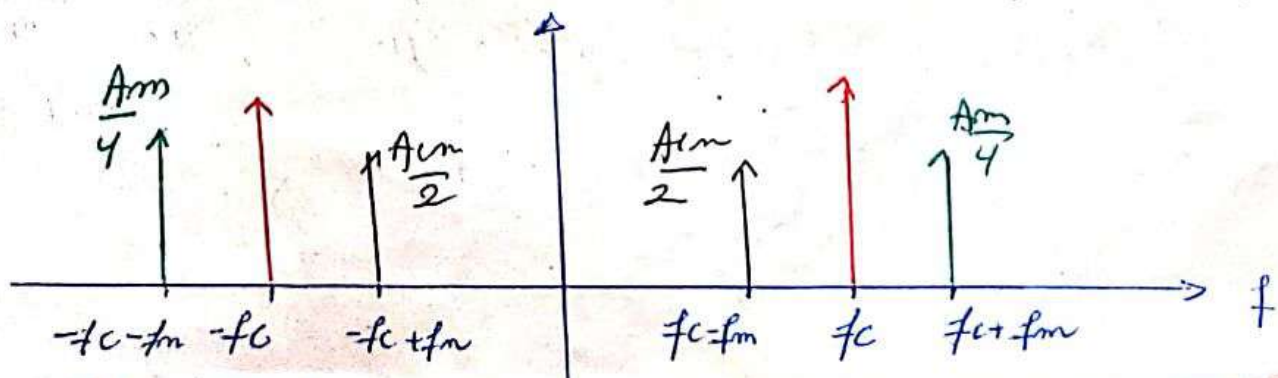


spectre de g(t).

$$G(f) = TF(g(t))$$

$$= \frac{A_c}{2} (\delta(f - f_c) + \delta(f + f_c)) + \frac{A_c m}{4} (\delta(f - (f_c + f_m)) + \delta(f + (f_c - f_m)) +$$

$$\frac{A_c m}{4} (\delta(f - (f_c + f_m)) + \delta(f - (f_c - f_m)))$$



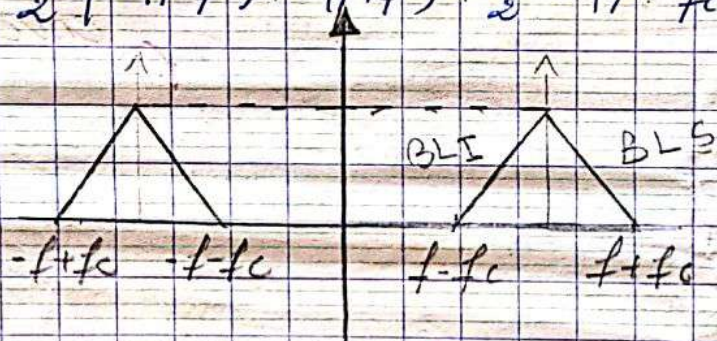




$$g(t) = A_c \cos 2\pi f_c t + k m(t) \cos 2\pi f_c t$$

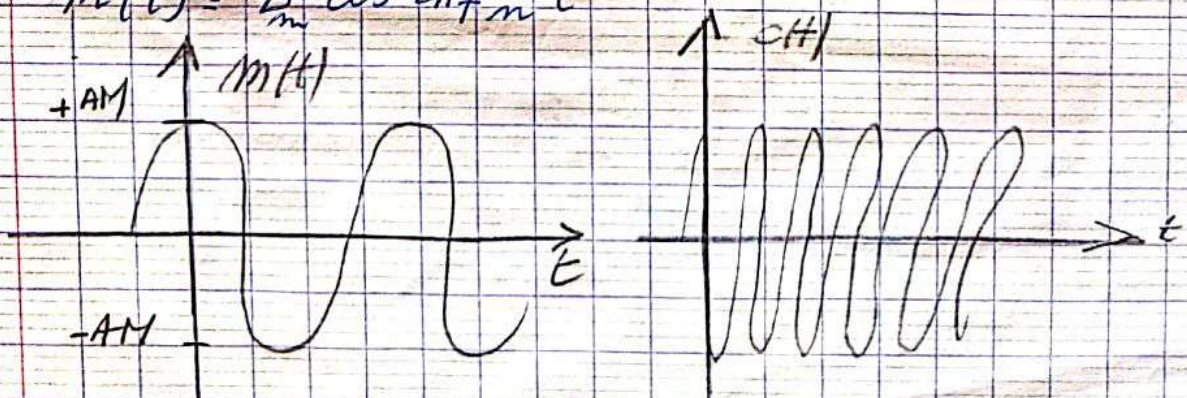
$$G(f) = \frac{A_c}{2} [\delta(f-f_c) + \delta(f+f_c)] + k M(f) * \frac{1}{2} [\delta(f-f_c) + \delta(f+f_c)]$$

$$= \frac{A_c}{2} [\delta(f-f_c) + \delta(f+f_c)] + \frac{k}{2} M(f-f_c) + \frac{k}{2} M(f+f_c)$$



Alors d'un signal AM.

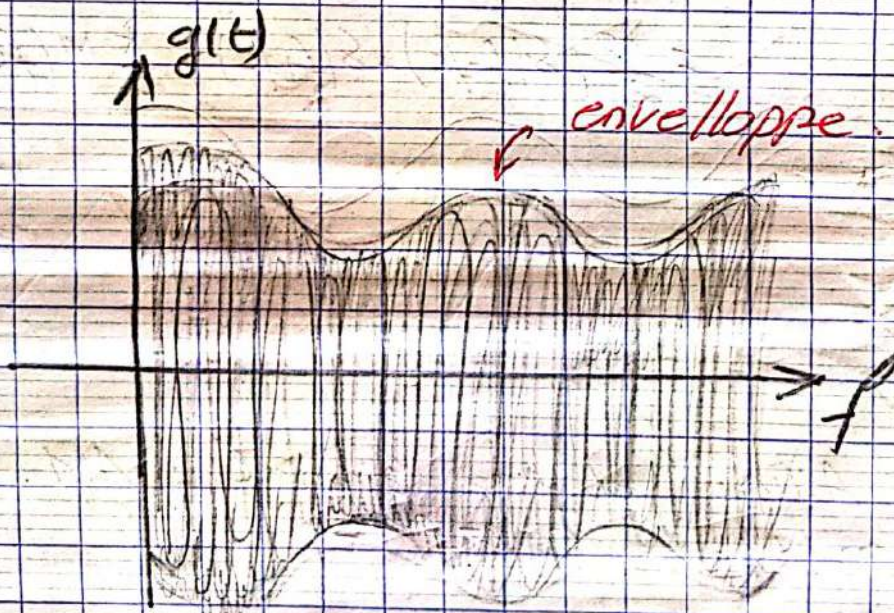
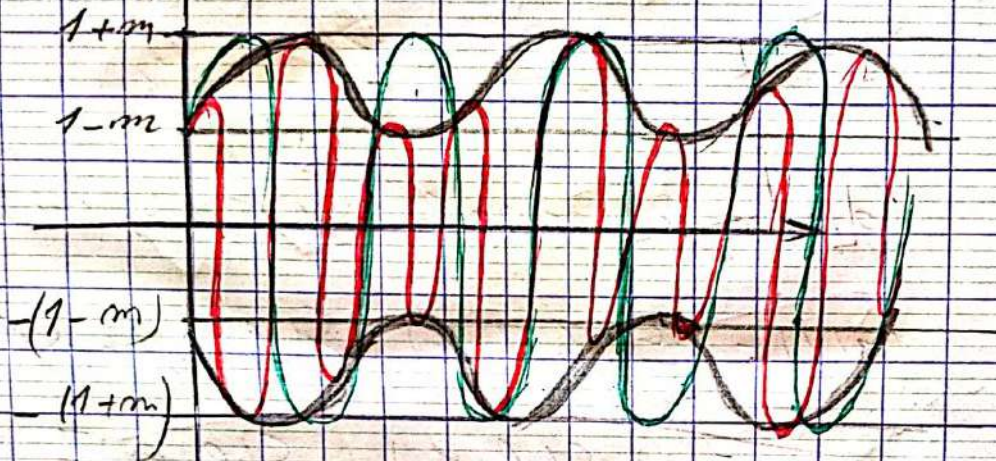
$$m(t) = D_m \cos 2\pi f_m t$$





$$g(t) = A_c \underbrace{(1 + m \cos 2\pi f_m t)}_{[-1, 1]} \underbrace{\cos 2\pi f_c t}_{[-1, 1]}$$

$$g(t) = [1 - m, 1 + m] \times [-1, 1]$$





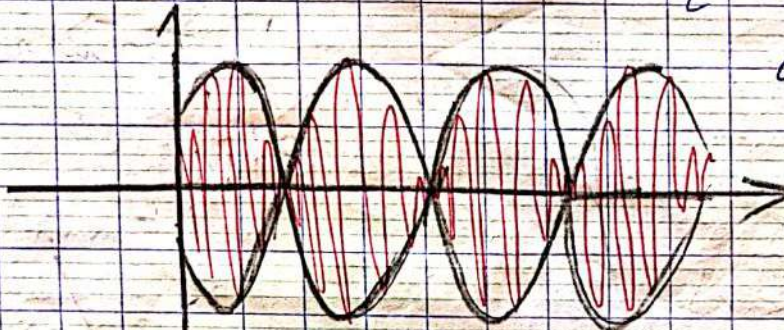
taux de modulation

$[1-m, 1+m]$

si  $m = 1$

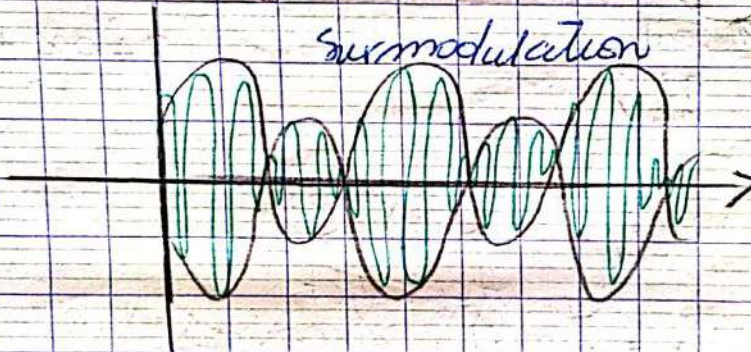
$$g(t) = A_c [1 + \cos 2\pi f_m t] \cos 2\pi f_c t$$

cos lente



si  $m > 1$

surmodulation



Calcul de la puissance :

$$P = \frac{V^2}{R}$$

cos  $\rightarrow \frac{V_{eff}^2}{R}, V_{eff} = \frac{V_{max}}{\sqrt{2}}$

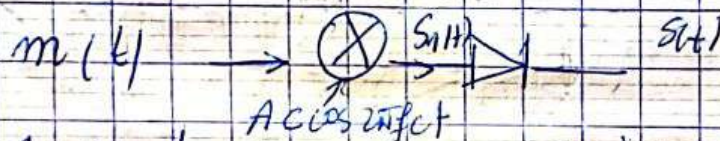
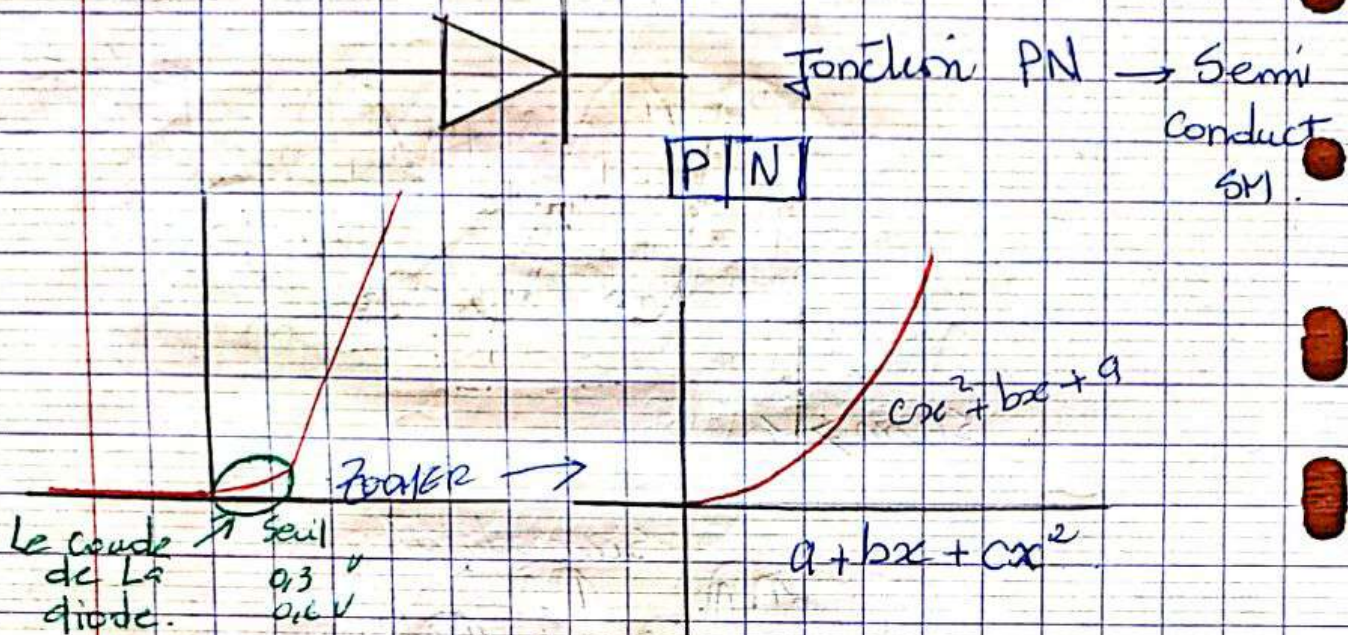
$$(f_c - f_m) \rightarrow \left( \frac{m A_M}{2\sqrt{2}} \right)^2 \frac{1}{R} = \frac{m^2 A_M^2}{8R}$$

$$(f_c + f_m) \rightarrow \left( \frac{m A_M}{2\sqrt{2}} \right)^2 \frac{1}{R} = \frac{m^2 A_M^2}{8R}$$

$$f_c \rightarrow \left( \frac{A_c}{\sqrt{2}} \right)^2 \frac{1}{R} = \frac{A_c^2}{2R}$$



# Génération d'un Signal AM.



$$S_1(t) = m(t) + A \cos 2\pi fct$$

$$S_1(t) = a + bx + cx^2$$

$$= a + b(m(t) + A \cos 2\pi fct) + c(m(t) + A \cos 2\pi fct)^2$$

$$\cos^2 x = \frac{1 + \cos 2x}{2}$$

$$S_1(t) = a + b m(t) + b A \cos 2\pi fct + c m^2(t) + c A^2 \cos^2 2\pi fct + 2c m(t) A \cos 2\pi fct$$

A

$$S_1(t) = \frac{CA^2}{2} + a + b(m(t) + A \cos 2\pi fct) + c(m(t) + A \cos 2\pi fct)^2 + \frac{bAC \cos 2\pi fct}{2} + \frac{CA^2 \cos 4\pi fct}{2}$$

$$\frac{1 + \cos 2x}{2} = \frac{CA^2}{2} + c \frac{A^2 \cos 4\pi fct}{2}$$

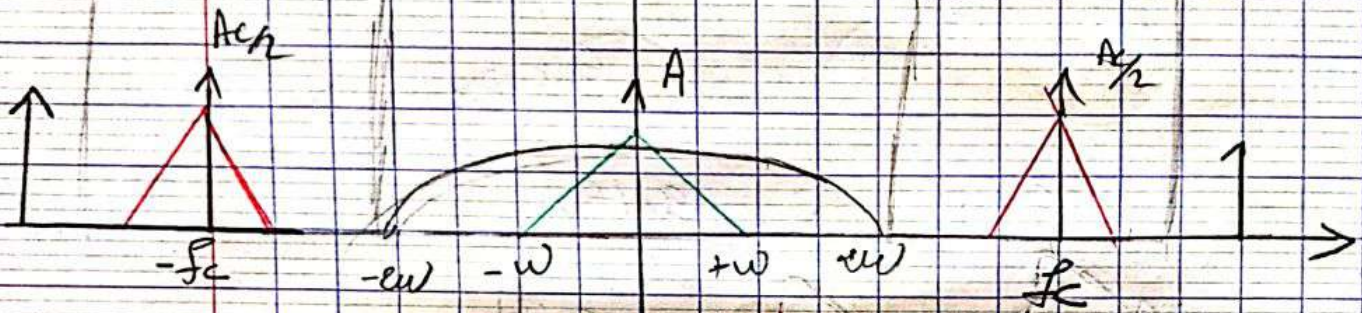
$$c A^2 \cos^2 2\pi fct$$

$$= bAC \cos 2\pi fct \left(1 + \frac{2c}{b} m(t)\right)$$

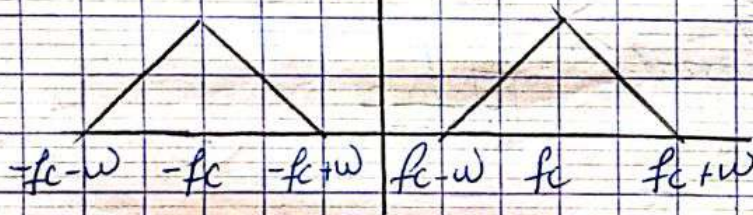
$$= \cos 2\pi fct = (bAC + 2AC m(t))$$



$$S(f) = \frac{1}{4} [A S(f) + b M(f) + c M_2(f) + \frac{bAc}{c} [S(f-f_c) + S(f+f_c)] + \frac{cAc}{4} [M(f-f_c) + M(f+f_c)] + \frac{CA^2}{4} [S(f-2f_c) + S(f+2f_c)]]$$



Après filtrage :

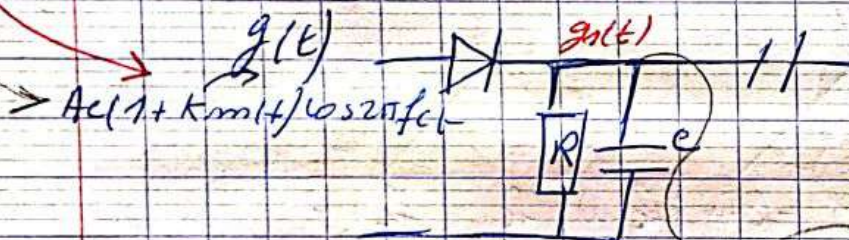
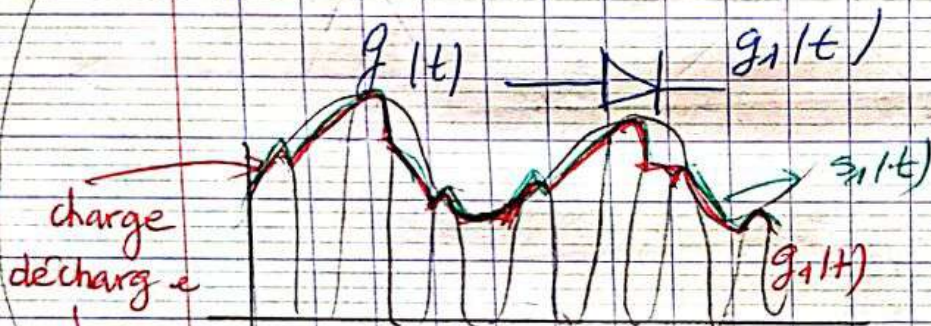
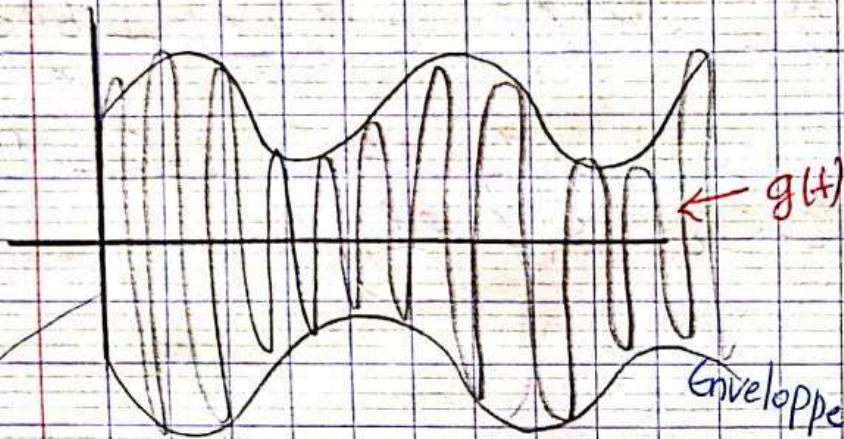


$$g(t) = A_c [1 + k_m(t)] \cos \pi f_c t$$

Demodulation of an signal AM

$$g(t) \rightarrow \boxed{\text{Demod}} \rightarrow m(t)$$





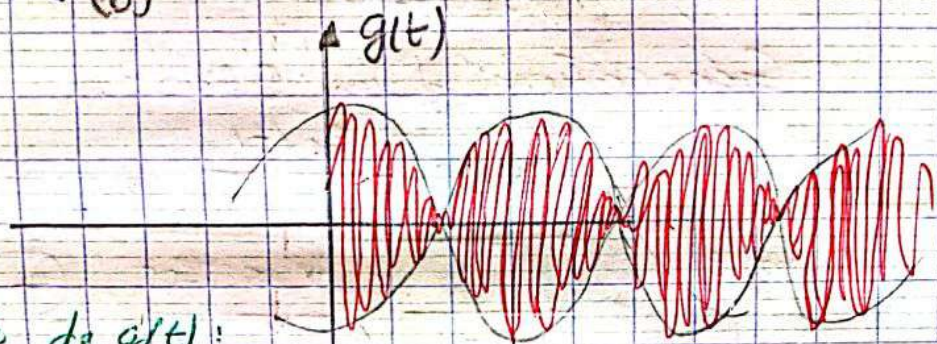
+ juste  
 -||- élimine  
 la composante  
 continue.



## Chapitre 03 : Modulation d'amplitude sans Porteuse.

$k$  (Gain de mod)

$$K_m(t) \rightarrow \text{Ac} \cos 2\pi f_c t \rightarrow g(t) = K_m(t) \cdot A_c \cos 2\pi f_c t$$



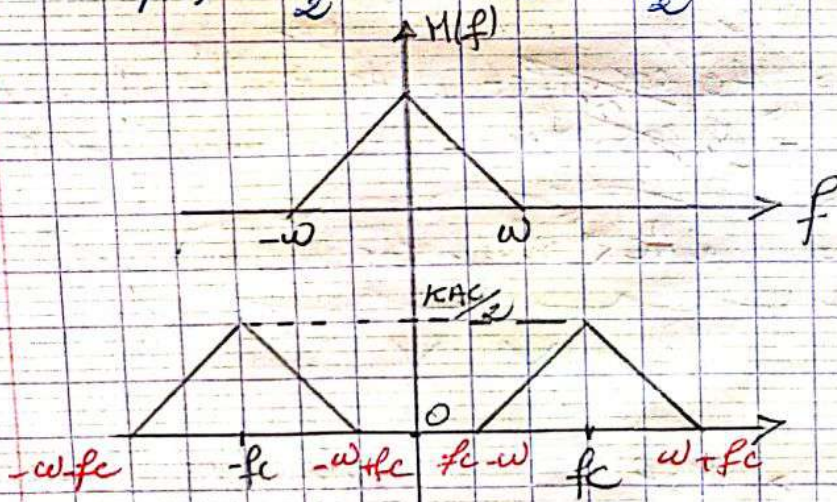
spectre de  $g(t)$ :

$$g(t) = K A_c m(t) \cdot \cos 2\pi f_c t$$

$$G(f) = K A_c M(f) \cdot \frac{1}{2} [\delta(f-f_c) + \delta(f+f_c)]$$

$$G(f) = K \frac{A_c}{2} [M(f-f_c) + M(f+f_c)]$$

$$G(f) = \frac{K A_c}{2} M(f-f_c) + \frac{K A_c}{2} M(f+f_c)$$





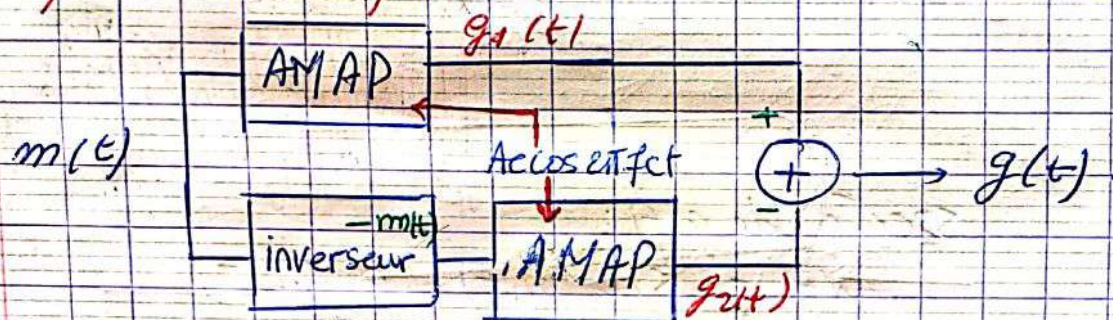
forme canonique :

$$g(t) = K A_c m(t) \cdot \cos 2\pi f_c t$$

$$g(t) = g_c(t) \cos 2\pi f_c t - g_s(t) \sin 2\pi f_c t$$

$$\Rightarrow \begin{cases} g_c(t) = K A_c m(t) \\ g_s(t) = 0 \end{cases}$$

Generation d'un SAMP.



$$g(t) = g_1(t) + g_2(t)$$

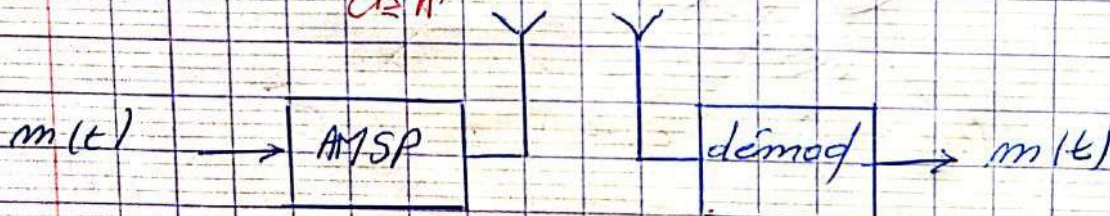
$$g_1(t) = A_c (1 + K m(t)) \cos 2\pi f_c t$$

$$g_2(t) = A_c (1 - K m(t)) \cos 2\pi f_c t$$

$$\hookrightarrow g(t) = \cancel{A_c \cos 2\pi f_c t} + A_c K m(t) \cos 2\pi f_c t - \cancel{A_c \cos 2\pi f_c t} + A_c K m(t) \cos 2\pi f_c t$$

$$g(t) = \boxed{2 A_c K m(t) \cos 2\pi f_c t} \quad \checkmark$$

est A'





$$\cos^2 x = \frac{1 + \cos 2x}{2}$$

$$g(t) = A_c m(t) \cos 2\pi f_c t \rightarrow \otimes \rightarrow s(t)$$

$$\uparrow A_c \cos 2\pi f_c t$$

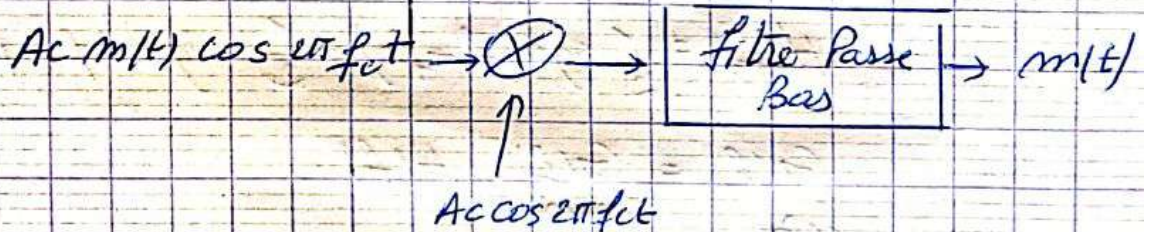
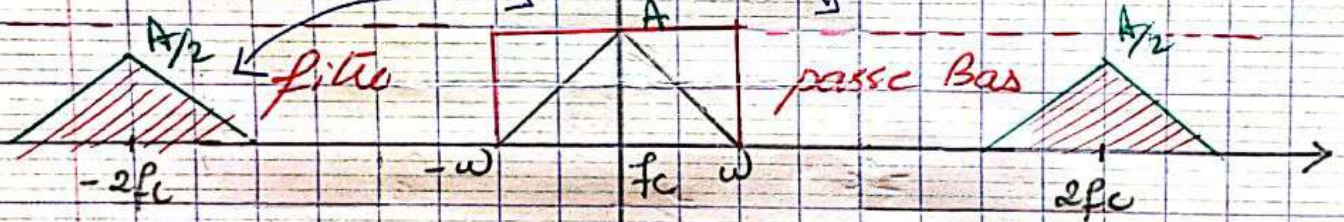
$$s(t) = A_c A_c' m(t) \cos^2 2\pi f_c t$$

$$= \frac{A_c A_c'}{2} m(t) (1 + \cos 4\pi f_c t)$$

$$= \frac{A_c A_c'}{2} m(t) + \frac{A_c A_c'}{2} m(t) \cos 4\pi f_c t$$

$$s(t) = A m(t) + A m(t) \cos 4\pi f_c t$$

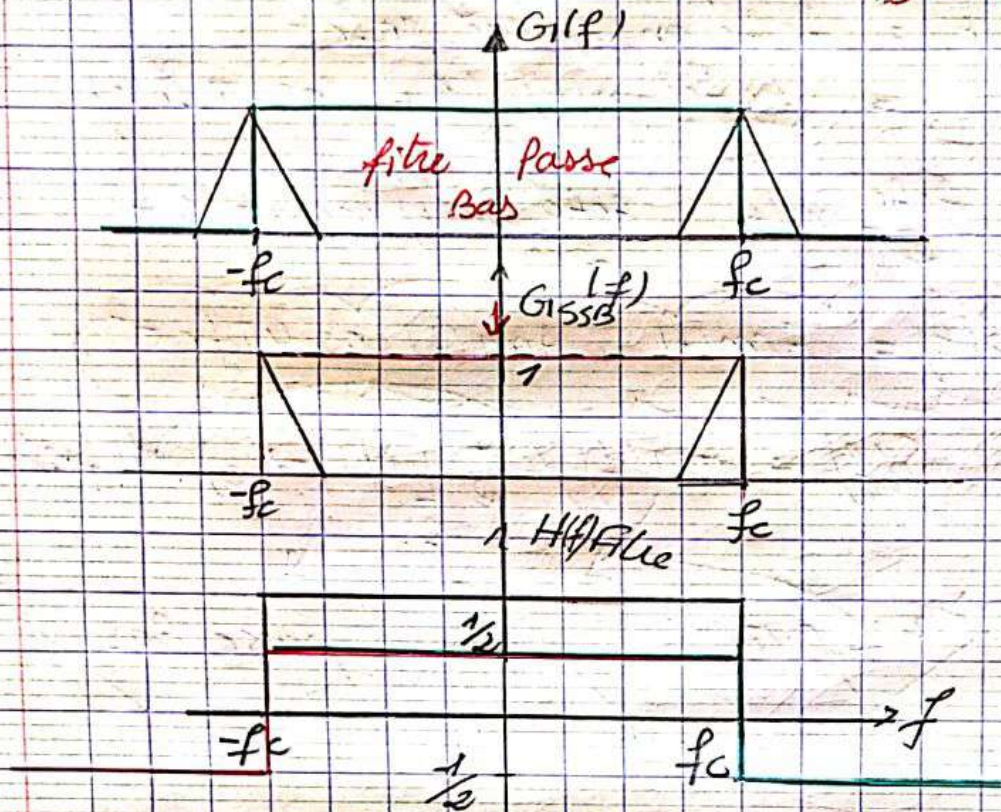
$$S(f) = A M(f) + \frac{A}{2} [M(f - 2f_c) + M(f + 2f_c)]$$





# Chapitre 04: Modulation à Bande latérale unique: BLU:--

SSB: Single Side Band



$$H(f) = \frac{1}{2} \operatorname{sgn}(f+fc) - \frac{1}{2} \operatorname{sgn}(f-fc)$$

$$G_{SSB}(f) = G_{AMSP}(f) \cdot H(f)$$

$$= G_{AMSP}(f) \cdot \left( \frac{1}{2} \operatorname{sgn}(f+fc) - \frac{1}{2} \operatorname{sgn}(f-fc) \right)$$

$$\begin{aligned} G_{SSB}(f) &= \left[ \frac{AC}{2} M(f-fc) + \frac{AC}{2} M(f+fc) \right] \cdot \left[ \frac{1}{2} \operatorname{sgn}(f+fc) - \frac{1}{2} \operatorname{sgn}(f-fc) \right] \\ &= \frac{AC}{4} M(f-fc) \cdot \operatorname{sgn}(f+fc) - \frac{AC}{4} M(f-fc) \cdot \operatorname{sgn}(f-fc) + \frac{AC}{4} M(f+fc) \cdot \operatorname{sgn}(f+fc) - \frac{AC}{4} M(f+fc) \cdot \operatorname{sgn}(f-fc) \end{aligned}$$

ona:  $\frac{AC}{2} M(f-fc) \cdot \frac{1}{2} \operatorname{sgn}(f+fc) = \frac{AC}{4} M(f-fc)$

pour simplifier



$$\textcircled{I} = \frac{Ac}{4} (M(f+f_c) \operatorname{sgn}(f+f_c)) = \frac{Ac}{4j} \cdot j (f+f_c) \operatorname{sgn}(f+f_c)$$

$$= -\frac{Ac}{4j} (-j M(f+f_c) \operatorname{sgn}(f+f_c))$$

$$= -\frac{Ac}{j4} \operatorname{TF} [TH[X(t)]]$$

$$\operatorname{TF}^{-1} [X(f+f_c)] = x(t) e^{-j2\pi f_c t}$$

$$\operatorname{TF}^{-1} [I] = -\frac{Ac}{j4} \hat{m}(t) e^{-j2\pi f_c t}$$

$$\textcircled{II} = -\frac{Ac}{4} (M(f-f_c) \operatorname{sgn}(f-f_c))$$

$$= -\frac{Ac}{j4} (-j M(f-f_c) \operatorname{sgn}(f-f_c))$$

$$\operatorname{TF}^{-1} [II] = \frac{Ac}{j4} \hat{m}(t) e^{j2\pi f_c t}$$

$$G_{SSB}(f) = \frac{Ac}{4} (M(f-f_c) + M(f+f_c)) + I + II$$

$$g_{SSB}(t) = \frac{Ac}{2} m(t) \cos 2\pi f_c t - \frac{Ac}{j4} \hat{m}(t) e^{-j2\pi f_c t}$$

$$+ \frac{Ac}{j4} \hat{m}(t) e^{j2\pi f_c t}$$

on développe expo:

$$= \frac{Ac}{2} m(t) \cos 2\pi f_c t - \frac{Ac}{j4} \hat{m}(t) \cos 2\pi f_c t - j \hat{m}(t) \sin 2\pi f_c t$$

$$+ \frac{Ac}{j4} \hat{m}(t) \cos 2\pi f_c t + j \hat{m}(t) \sin 2\pi f_c t$$

$$= \frac{Ac}{2} m(t) \cos 2\pi f_c t + \frac{Ac}{2} \hat{m}(t) \sin 2\pi f_c t$$



à retenir

Forme canonique

$$g(t) = \frac{AC}{2} m(t) \cos 2\pi f_c t + \frac{AC}{2} \hat{m}(t) \sin 2\pi f_c t$$

$$\begin{cases} g_c(t) = \frac{AC}{2} m(t) \\ g_s(t) = -\frac{AC}{2} \hat{m}(t) \end{cases}$$

Exemple :

$$m(t) = A_m \cos 2\pi f_m t$$

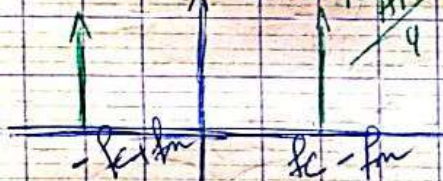
$$M(f) = \frac{A_m}{2} (\delta(f - f_m) + \delta(f + f_m))$$



$$g(t) = \frac{AC}{2} m(t) \cos 2\pi f_c t + \frac{AC}{2} \hat{m}(t) \sin 2\pi f_c t$$

$$= \frac{A_m A_c}{2} \cos 2\pi f_m t \cos 2\pi f_c t + \frac{A_m A_c}{2} \sin 2\pi f_m t \sin 2\pi f_c t$$

SSB  
Inférieur  
supérieur

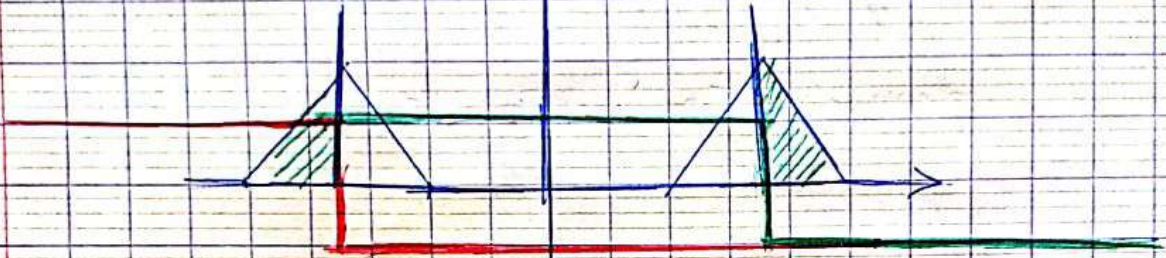


(On envoie qu'une seule bande du signal)

Génération du signal SSB :



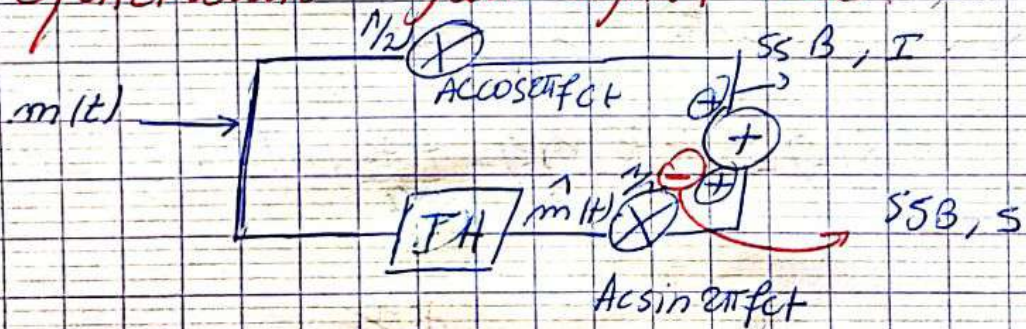
SSB - Symétrique → Filtrage passe haut.



$$H(f) = -\frac{1}{2} \operatorname{sgn}(f + f_c) - \frac{1}{2} \operatorname{sgn}(f - f_c)$$

$$g_{SSB, S}(t) = \frac{A_c}{2} m(t) \cos 2\pi f_c t - \frac{A_c}{2} \hat{m}(t) \sin 2\pi f_c t$$

Génération d'un signal SSB





## Demodulation AMSP:

$$g(t) = A_c m(t) \cos 2\pi f_c t$$

démultiplexage cohérent

$$Y \left[ g(t) \right] \xrightarrow{\otimes} S(t) = g(t) \cdot A_c' \cos 2\pi f_c t$$

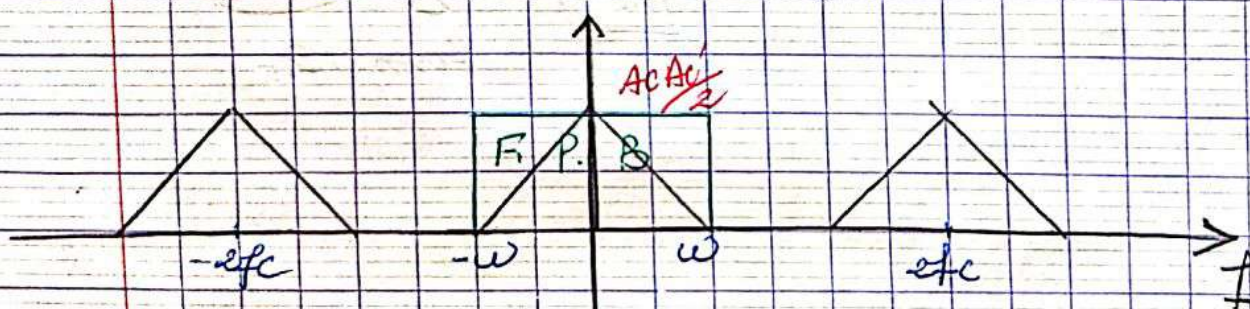
$$S(t) = A_c A_c' m(t) \cos^2 2\pi f_c t$$

$$S(t) = A_c A_c' m(t) \left( \frac{1 + \cos 4\pi f_c t}{2} \right)$$

$$S(t) = \frac{A_c A_c'}{2} m(t) + \frac{A_c A_c'}{2} m(t) \cos 4\pi f_c t$$

Spectre de  $S(t)$ :

$$S(f) = \frac{A_c A_c'}{2} M(f) + \frac{A_c A_c'}{4} (M(f - 2f_c) + M(f + 2f_c))$$



$$A_c m(t) \cos 2\pi f_c t \xrightarrow{\otimes} \text{FPB} \rightarrow A_c' m(t)$$

$$A_c' \cos 2\pi f_c t + \phi$$

↳ phase (constant)



Demoq  
SSB-I :

\*\*\*

$$g(t) \rightarrow s(t) = g(t) \cdot A_c' \cos 2\pi f_c t$$

$$A_c' \cos 2\pi f_c t$$

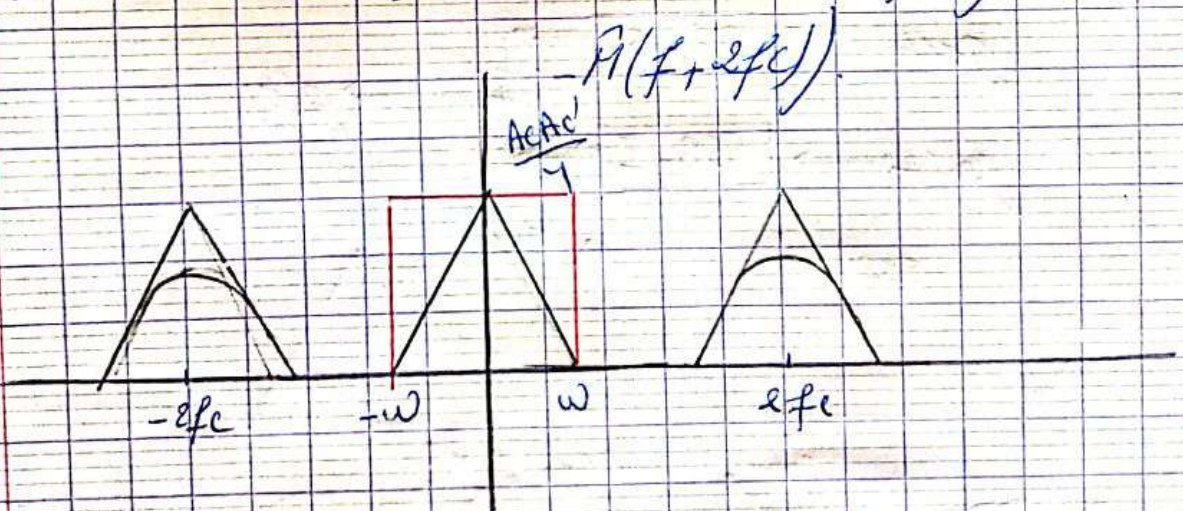
$$s(t) = \frac{A_c A_c'}{2} m(t) \cos 2\pi f_c t + \frac{A_c A_c'}{2} \hat{m}(t)$$

$$s(t) = \frac{A_c A_c'}{4} m(t) (1 + \cos 4\pi f_c t) + \frac{A_c A_c'}{4} \hat{m}(t) \sin 4\pi f_c t$$

$$= \frac{A_c A_c'}{4} m(t) + \frac{A_c A_c'}{4} m(t) \cos 4\pi f_c t + \frac{A_c A_c'}{4} \hat{m}(t) \sin 4\pi f_c t$$

Spectre de  $S(f)$ :

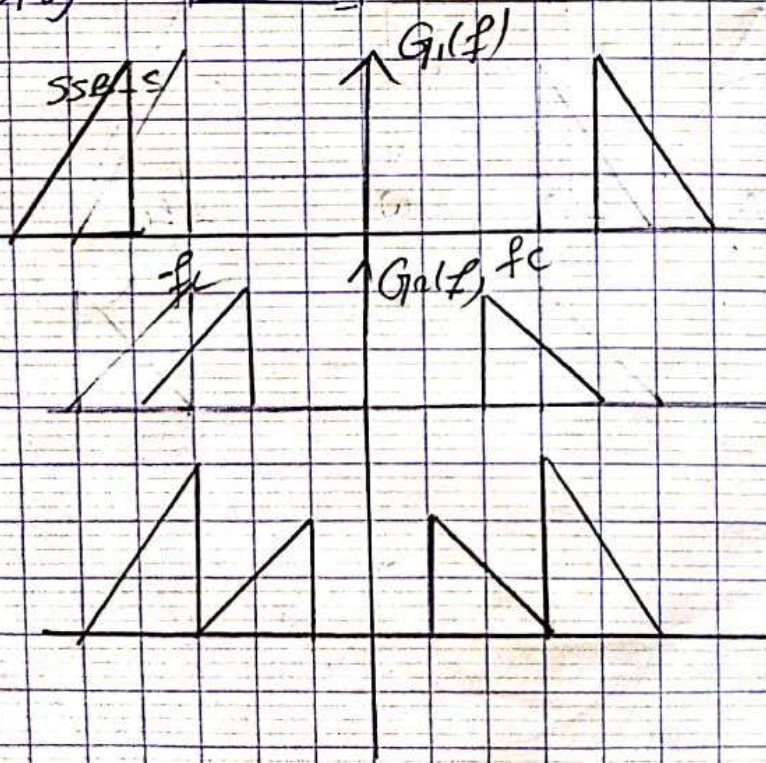
$$S(f) = \frac{A_c A_c'}{2} M(f) + \frac{A_c A_c'}{8} M(f - 2f_c) + M(f + 2f_c) + \frac{A_c A_c'}{j8} (\hat{M}(f - 2f_c) - \hat{M}(f + 2f_c))$$





# Modulation bande latérale Indépendantes BLI

ADN p334

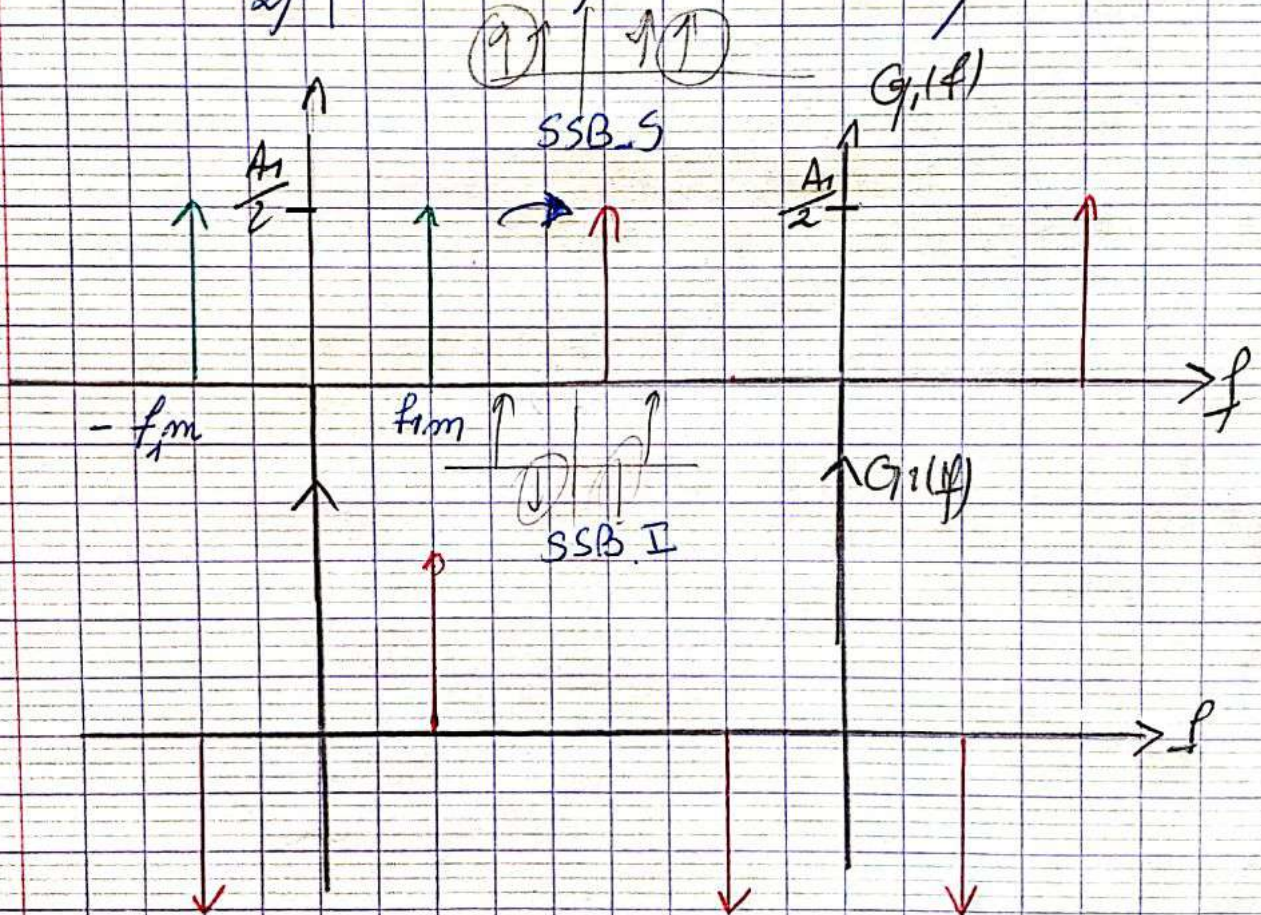




$$\left\{ \begin{array}{l} m_1(t) = A_1 \cos 2\pi f_m t \rightarrow \text{SSB-S} \\ m_2(t) = A_2 \sin 2\pi f_m t \rightarrow \text{SSB-I} \end{array} \right.$$

$$M_1(f) = \frac{1}{2} (\delta(f - f_m) + \delta(f + f_m))$$

$$M_2(f) = \frac{1}{2j} (\delta(f - f_m) - \delta(f + f_m))$$

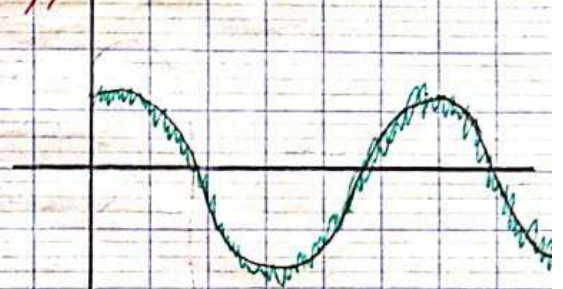




14-01-2019

## Demodulation

### ① detection d'enveloppe

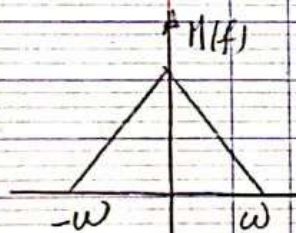


### ② demodulation synchrone

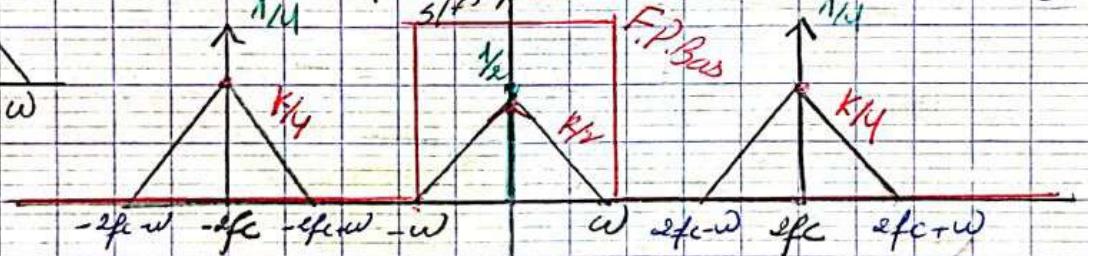
$$g(t) = A_c(1 + k_m(t)) \cos 2\pi f_c t \quad \times \quad A_c' \cos 2\pi f_c t \quad \rightarrow \quad s(t) = A_c A_c' (1 + k_m(t)) \cos^2 2\pi f_c t$$

$$A_c' \cos 2\pi f_c t$$

$$s(t) = A_c A_c' (1 + k_m(t)) \left( \frac{1 + \cos 4\pi f_c t}{2} \right)$$

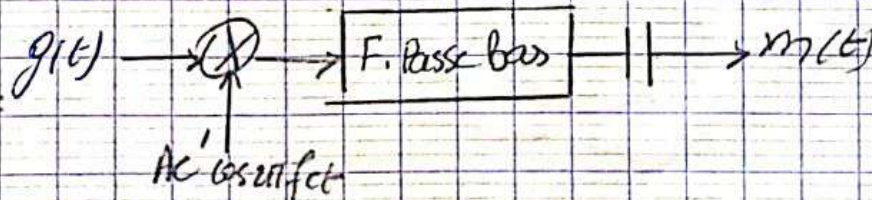


$$= A_c A_c' \left( \frac{1}{2} + k_m(t) \right) \frac{\cos 4\pi f_c t}{2} + k_m(t) \frac{\cos 4\pi f_c t}{2}$$



pour  
eliminer

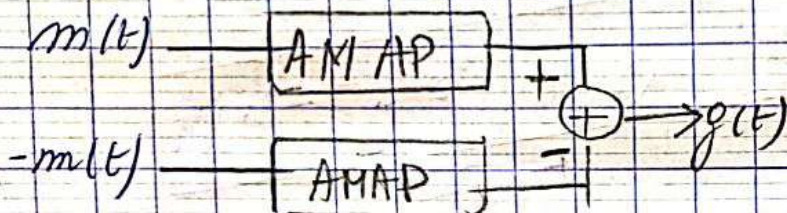
la composante  
continue on doit  
ajouter un  $\frac{1}{2}$





modulation AM sans porteur

$$g(t) = K A_c m(t) \cos 2\pi f_c t$$



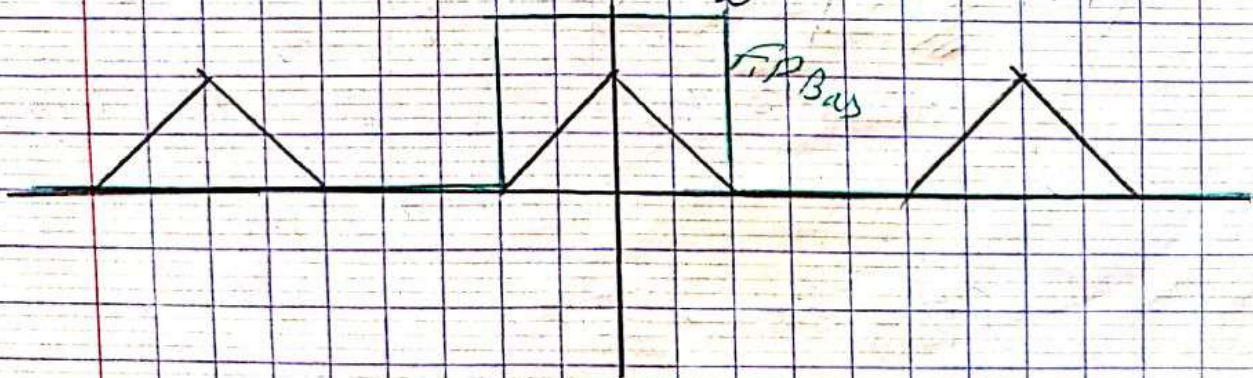
$$\begin{aligned} g(t) &= A_c (1 + K m(t)) \cos 2\pi f_c t - A_c (1 - K m(t)) \cos 2\pi f_c t \\ &= A_c \cos 2\pi f_c t + A_c K m(t) \cos 2\pi f_c t + A_c \cos 2\pi f_c t + A_c K m(t) \cos 2\pi f_c t \end{aligned}$$

$$g(t) = 2K A_c m(t) \cos 2\pi f_c t$$

démodulation cohérente synchrone:

$$g(t) \rightarrow \otimes \xrightarrow{A_c' \cos 2\pi f_c t}$$

$$\begin{aligned} s(t) &= A_c A_c' m(t) \cos^2 2\pi f_c t = \frac{A_c A_c'}{2} m(t) (1 + \cos 4\pi f_c t) \\ &= \frac{A_c A_c'}{2} m(t) + \frac{A_c A_c'}{2} m(t) \cos 4\pi f_c t \end{aligned}$$

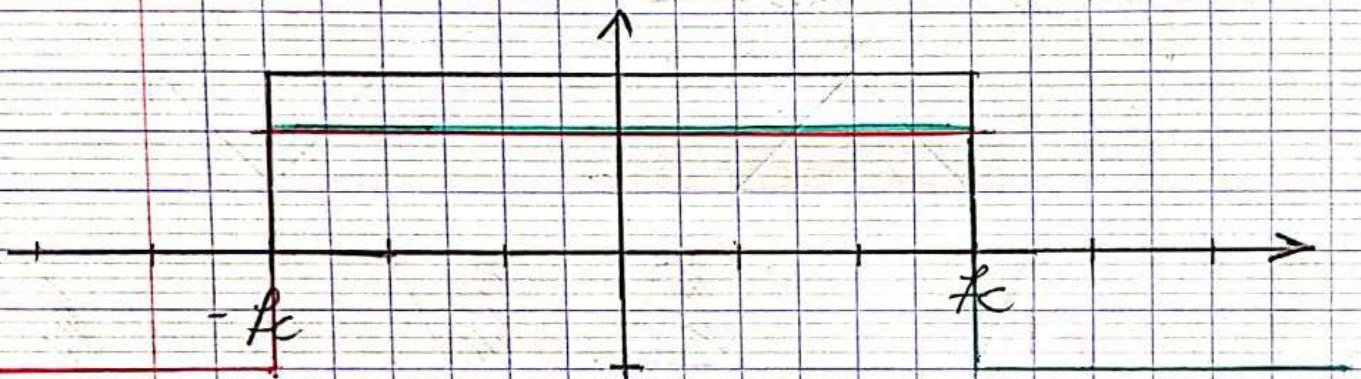




SSB - Bande latérale unique:



$$G_{SSB}(f) = H(f) S(f)$$



$$H(f) = \frac{1}{2} \operatorname{sgn}(f + f_c) - \frac{1}{2} \operatorname{sgn}(f - f_c)$$

$$G(f) = \left( \frac{A_c}{2} |M(f - f_c) + M(f + f_c)| \right) \left( \frac{1}{2} (\operatorname{sgn}(f + f_c) - \operatorname{sgn}(f - f_c)) \right)$$

Hilbert  $g(t) = \frac{A_c}{2} m(t) \cos 2\pi f_c t + \frac{A_c}{2} \hat{m}(t) \sin 2\pi f_c t$

opération :

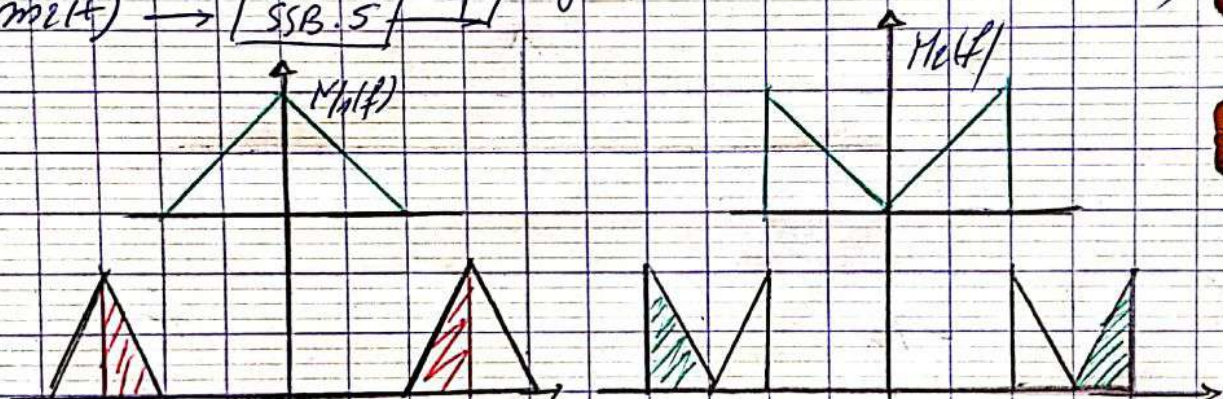




Baseband lateral independent, BLI  
 2 msg different  $m_1(t)$  et  $m_2(t)$

$m_1(t) \rightarrow \text{BLI}$   
 $m_2(t) \rightarrow \text{BLS}$

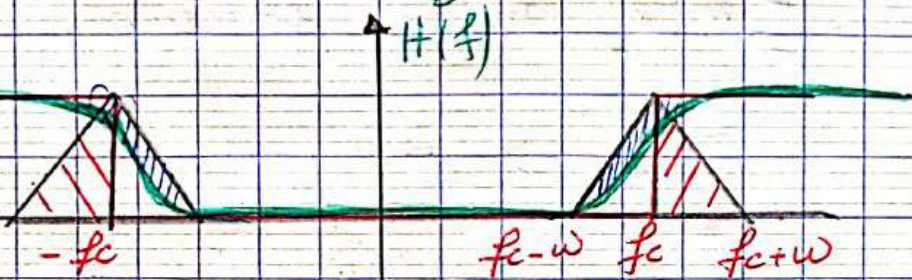
$m_1(t) \rightarrow \boxed{\text{SSB-LI}}$   
 $m_2(t) \rightarrow \boxed{\text{SSB-S}}$   
 $\oplus \Rightarrow f(t) \Rightarrow G(f) = G_1(f) + G_2(f)$





# Chapitre 6 : Modulation à bande latérale atténuée : Vestigial side Band VSB

Le filtre en rouge n'est pas réalisable dans l'électronique analogique ce qui pousse à avoir le vest



VSB: télévision

condition de  $H(f)$ :

$$H(f-f_c) + H(f+f_c) = 1$$

$$G_{VSB}(f) = \frac{Ac}{2} (M(f-f_c) + M(f+f_c)) \cdot H(f)$$

$$G_c(f) = \begin{cases} G_{VSB}(f-f_c) + G_{VSB}(f+f_c) & |f| < w \\ 0 & |f| > w \end{cases}$$

$$G_{VSB}(f) = \frac{Ac}{2} (M(f-f_c) + M(f+f_c)) \cdot H(f)$$

on remplace dans

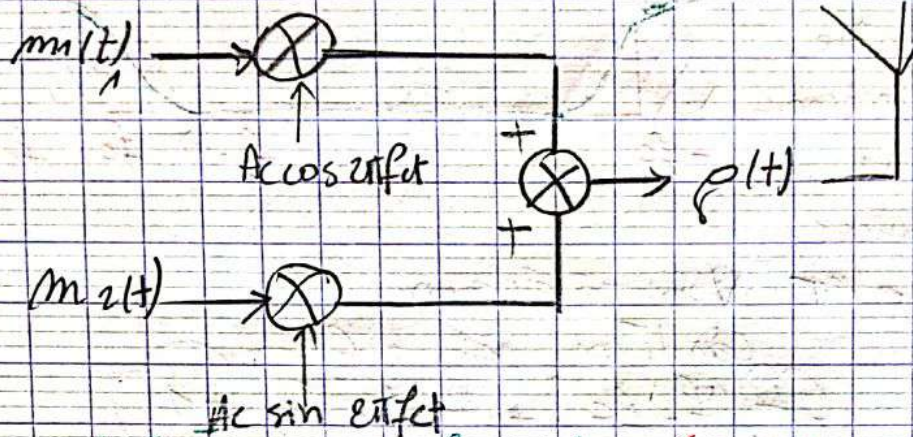
$$\begin{aligned} G_{VSB}(f) &= \frac{Ac}{2} (M(f-2f_c) + M(f)) \cdot H(f-f_c) \\ &\quad + \frac{Ac}{2} (M(f) + M(f+2f_c)) \cdot H(f+f_c) \\ &= \frac{Ac}{2} M(f) (H(f-f_c) + H(f+f_c)) + \\ &\quad \frac{Ac}{2} M(f-2f_c) H(f-f_c) + \frac{Ac}{2} M(f+2f_c) H(f+f_c) \end{aligned}$$

$$G_c(f) = \begin{cases} \frac{Ac}{2} M(f) (H(f-f_c) + H(f+f_c)) & |f| < w \\ 0 & \text{ailleurs} \end{cases}$$



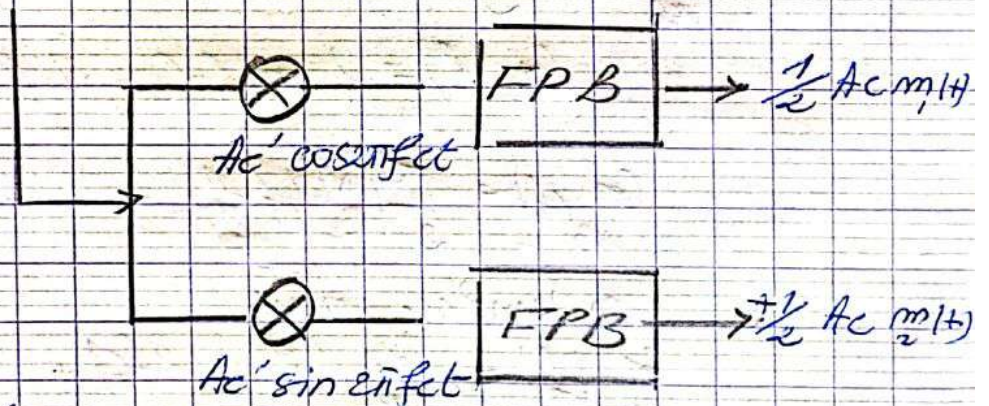
## Modulation d'amplitude en quadrature QAM.

On va transmettre deux msg  $m_1(t)$ ,  $m_2(t)$   
 $\rightarrow [-1, 1]$  même porteuse.



## Démodulation cohérente (synchrone).

$$p(t) = A_c m_1(t) \cos 2\pi f_c t + A_c m_2(t) \sin 2\pi f_c t.$$

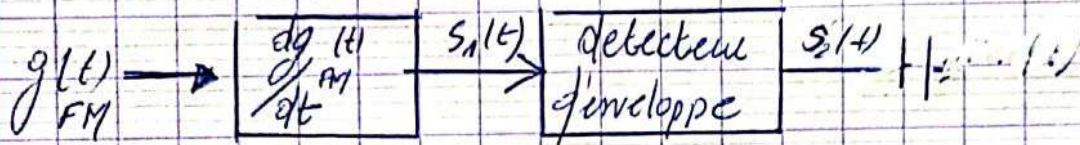


$$p_c(t) = A_c m_1(t)$$

$$p_s(t) = -A_c m_2(t)$$



## Démodulation FM par discrimination :

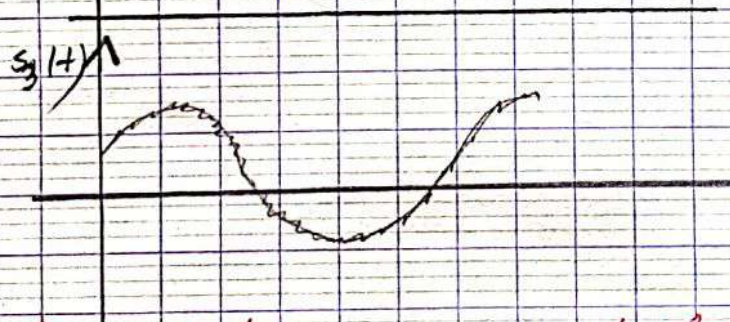


$$g_{FM}(t) = AC \cos \left( 2\pi f_c t + 2\pi k_f \int m(t) dt \right)$$

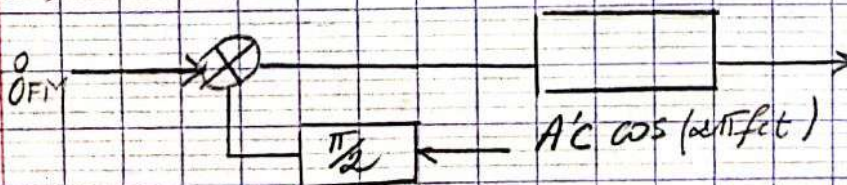
$$e_n(t) = \frac{dg_{FM}}{dt} = -AC \sin \left( 2\pi f_c t + 2\pi k_f \int m(t) dt \right) \cdot \left( 2\pi f_c + 2\pi k_f m(t) \right)$$

$$= -AC \cdot 2\pi f_c \left( 1 + \frac{k_f m(t)}{f_c} \right) \sin \left( 2\pi f_c t + 2\pi k_f \int m(t) dt \right)$$

$$= AC' \left( 1 + k_{FM} m(t) \right) \sin \left( 2\pi f_c t + 2\pi k_f \int m(t) dt \right)$$



## 2) Démodulateur en quadrature : (1/2)





$$G_c(f) = \frac{Ac}{2} M(f) \rightarrow g_c(t) = \frac{Ac}{2} m(t)$$

$$G_s(f) = j (G(f-f_c) - G(f+f_c))$$

$$G_s(f) = \frac{Ac}{2} (M(f-f_c) + M(f+f_c)) H(f)$$

$$G_s(f) = j \frac{Ac}{2} (M(f-f_c) + M(f)) \cdot H(f-f_c) - j \frac{Ac}{2} (M(f+2f_c) + M(f)) \cdot H(f+f_c)$$

$$= j \frac{Ac}{2} M(f) (H(f-f_c) + H(f+f_c)) + j \frac{Ac}{2} (M(f-f_c) - H(f-f_c)) - j \frac{Ac}{2} (M(f+2f_c) - H(f+f_c))$$

Du coup

$$G_s(f) = \begin{cases} j \frac{Ac}{2} M(f) (H(f-f_c) - H(f+f_c)) & \text{si } f \text{ est } \omega \\ 0 & \text{ailleurs} \end{cases}$$

$$\Rightarrow g_s(t) = \mathcal{F}^{-1}[G_s(f)] = m_s(t)$$

Donc

$$g_{vso}(t) = \frac{Ac}{2} m(t) \cos(2\pi f_c t) - \frac{Ac}{2} \sin(2\pi f_c t) m_s(t)$$



# Chapitre 06: Les composants d'une chaîne de transmission

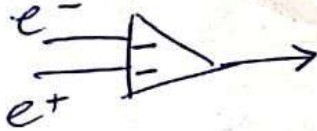
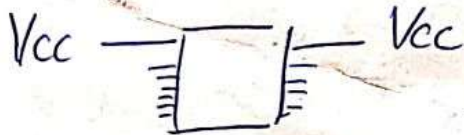
Amplification / Atténuation

Filtrage

- Génération de la porteuse

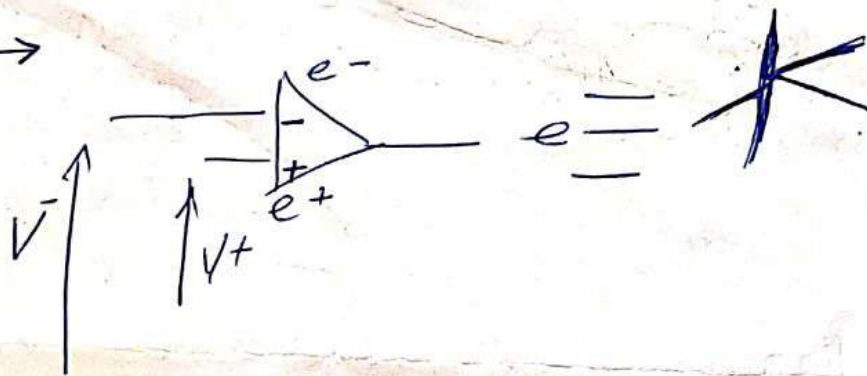
## L'amplification opérationnelle (Ampli op)

C'est un circuit intégré

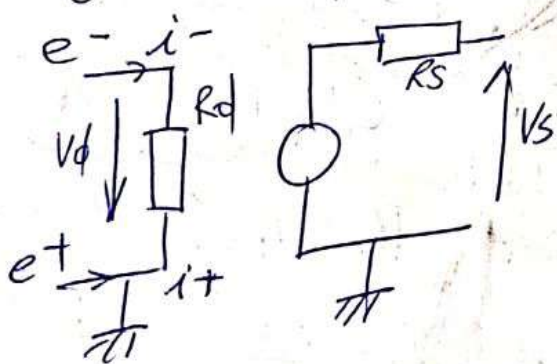


$e^+$  : entrée non inverseuse

$e^-$  : " inverseuse



Le circuit équivalent



$V_d = e^+ - e^-$  : tension différentielle

$A_d$  = gain

$R_d$  = Résistance d'entrée

$R_s$  = Résistance de sortie

Ampli idéal

$A_d \rightarrow$  très élevé ( $\infty$ )

$R_d \rightarrow$  très grande ( $\infty$ )

$i^+ = 0$

$V_d = 0$

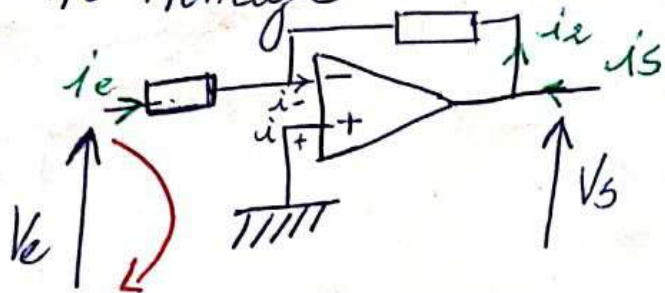
$i^- = 0$

$R_s$  (très petite (0-2))

①



# 1. Montage Inverseur :



Exemple :

$$V_e = 10 \text{ mV}$$

$$R_1 = 10 \text{ k}\Omega$$

$$R_2 = 1 \text{ M}\Omega$$

$$V_s = -1 \text{ V}$$

$$V_e - R_1 i_e + V_d = 0$$

$$V_e = R_1 i_e - V_d = R_1 i_e$$

$$i^- = i_2 + i_e = 0$$

$$i_e = -i_2$$

$$V_s - R_2 i_2 + V_d = 0$$

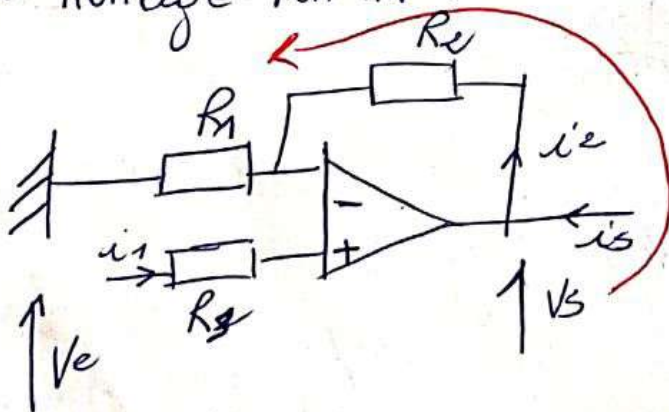
$$V_s = R_2 i_2$$

$$V_s = -R_2 i_e \Rightarrow V_s = -\frac{R_2}{R_1} V_e$$

$$\text{OR } V_e = R_1 i_e$$

$$\frac{V_s}{V_e} = -\frac{R_2}{R_1}$$

# 2. Montage non inverseur :



$$V_s = R_f i_2 - R_1 i_1$$

$$i^- = i_2 - i_1 \Rightarrow i_2 = i_1$$

$$V_s = (R_1 + R_f) i_2$$

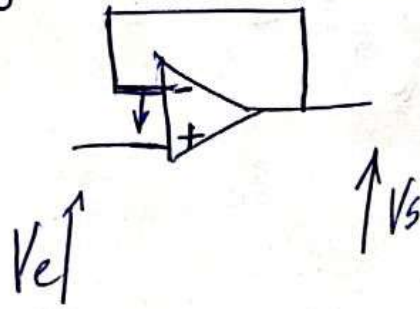
$$V_e = R_2 i_1 \quad V_e = R_2 i_2$$

$$V_s = \frac{R_1 + R_f}{R_2} V_e$$

$$\frac{V_s}{V_e} = \left(1 + \frac{R_f}{R_2}\right)$$



3- Montage Suiveur:



$$V_s = -V_d + V_e$$

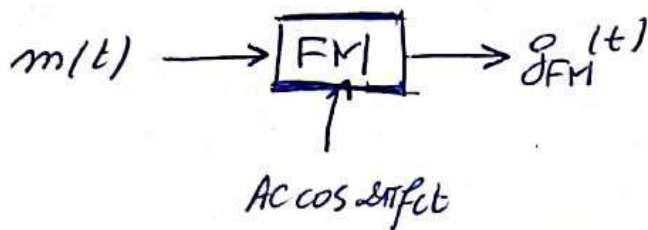
$$V_s = V_e$$

→ Utilisation

Adaptation d'impédance . ( Ampli suiveur  
impédance d'entrée  $\infty$  )



# Modulation de fréquence



$$g(t) = A \cos(2\pi f(t) t)$$

freq  
instantanée

la fréquence varie dans le temps,  $f(t)$  est message.

$$\cos(\theta) \rightarrow \cos(2\pi f_0 t) \quad f \text{ est constante}$$

$$\cos(\omega_0 t) \quad \omega_0 = 2\pi f_0$$

$$\theta = \int \omega(t) dt$$

si  $\omega = \omega_0 \rightarrow \omega(t) = \cos \int 2\pi f_0 dt = \cos(2\pi f_0 t)$

$m(t) \rightarrow$  portée par  $f_c$ .

$$f(t) = f_c + k_f m(t)$$

$k_f$  = sensibilité du modulateur

$$g(t) = A \cos(\theta(t))$$

$$= A \cos\left(\int 2\pi (f_c + k_f m(t)) dt\right)$$

$$= A \cos\left(\int 2\pi f_c dt + \int 2\pi k_f m(t) dt\right)$$

$$g(t) = A \cos(2\pi f_c t + 2\pi k_f \int m(t) dt)$$

FM

Cas particulier :  $m(t) = A_m \cos 2\pi f_m t$

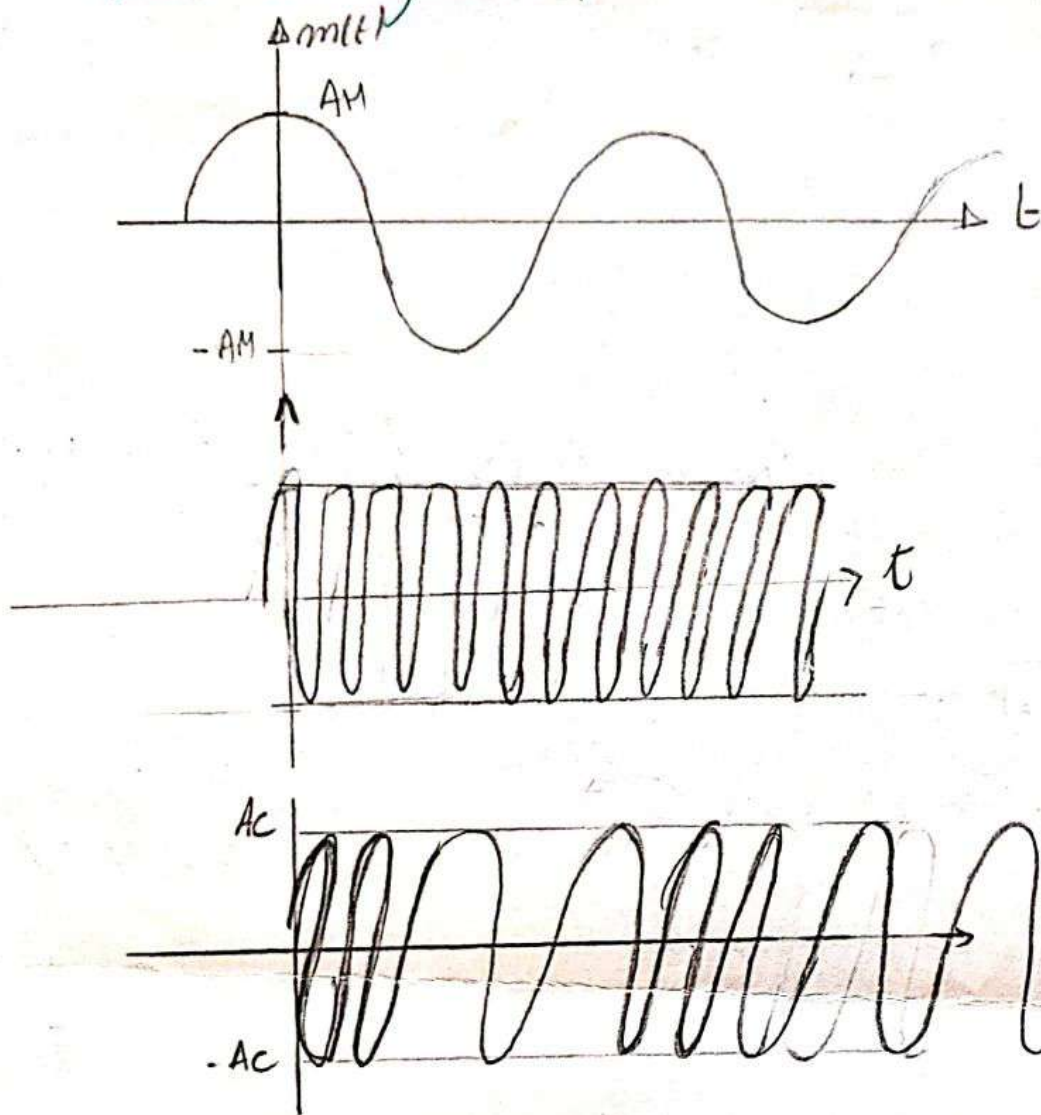
$$g(t) = A \cos(2\pi f_c t + 2\pi k_f \int A_m \cos 2\pi f_m t dt)$$

$$= A \cos(2\pi f_c t + \frac{2\pi k_f A_m}{2\pi f_m} \sin 2\pi f_m t)$$

$$g_{\text{FM}}(t) = A \cos(2\pi f_c t + \frac{k_f A_m}{f_m} \sin 2\pi f_m t)$$



Allure Du signal FM:



$$f(t) = f_c + k_f m(t) \rightarrow f_c + k_f A_m \cos 2\pi f_m t$$

$$= f_c + \Delta f m(t), \quad \Delta f = k_f A_m \text{ excursion en fr.}$$

$$f_c - \Delta f \leq f(t) \leq f_c + \Delta f.$$

$$g_{FM}^{(H)} = A_C \cos \left( 2\pi f_c t + \underbrace{\frac{\Delta f}{f_m}}_{\beta} \sin 2\pi f_m t \right).$$

$$g_{FM}^{(t)} = A_C \cos \left( \underbrace{2\pi f_c t}_a + \underbrace{\beta \sin 2\pi f_m t}_b \right)$$

indice de modulation

$$g_{FM}^{(t)} = A_C \cos(2\pi f_c t) \cdot \cos(\beta \sin 2\pi f_m t) - A_C \sin 2\pi f_c t \sin(\beta \sin 2\pi f_m t)$$



1<sup>er</sup> cas:  $\beta \ll 1$

mod FM à bande latérale étroite

$$\beta \cdot \sin 2\pi f_m t \ll 1.$$

si

$$\begin{cases} \sin x = x \\ x = \beta \sin 2\pi f_m t. \\ \cos x = 1 \end{cases}$$

developpement de Taylor.

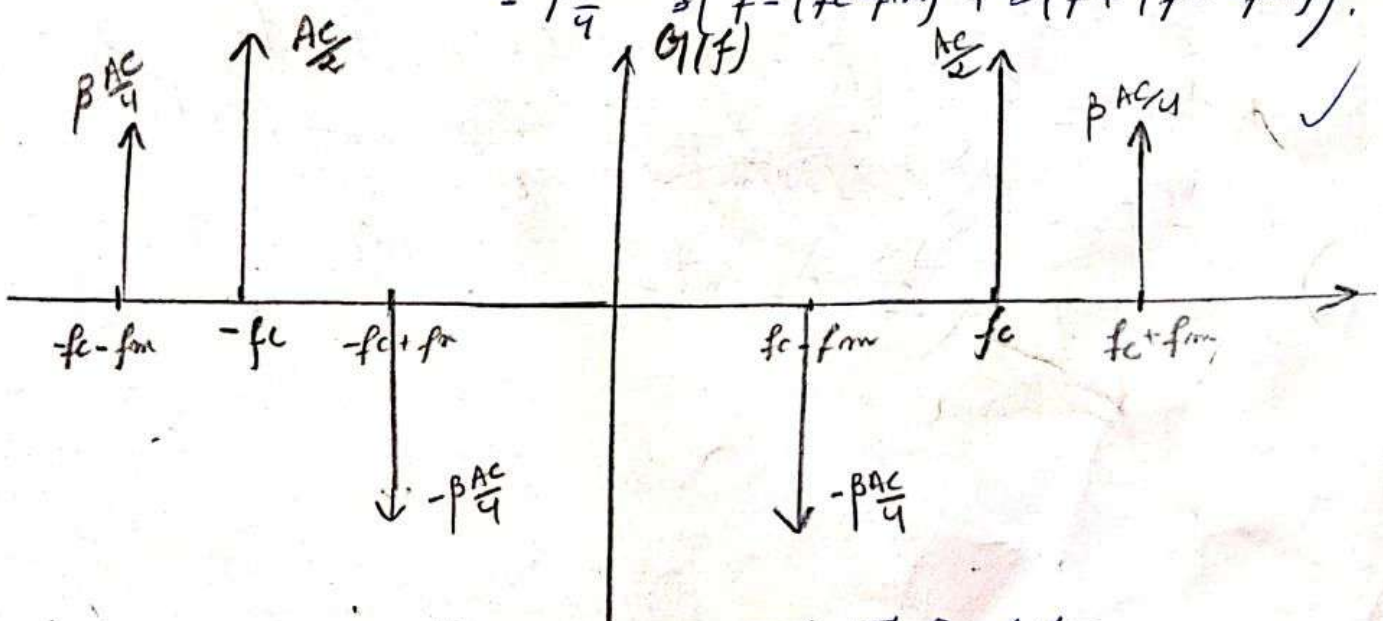
$$g(t)_{FM} = A_c \cos(2\pi f_c t) - A_c \beta \sin 2\pi f_c t \cdot \sin 2\pi f_m t$$

$$\Rightarrow g(t) = A_c \cos 2\pi f_c t$$

$$= \frac{\beta A_c}{2} (\cos 2\pi (f_c - f_m) t) - \cos(2\pi (f_c + f_m) t)$$

$$= A_c \cos 2\pi f_c t + \beta \frac{A_c}{2} \cos(2\pi (f_c + f_m) t) - \beta \frac{A_c}{2} \cos(2\pi (f_c - f_m) t)$$

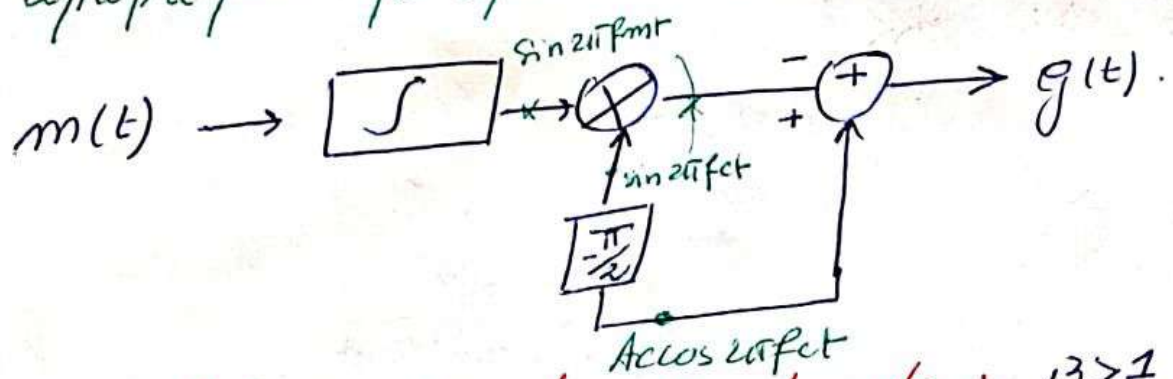
$$G(f) = \frac{A_c}{2} (\delta(f - f_c) + \delta(f + f_c)) + \frac{\beta A_c}{4} \delta(f - (f_c + f_m)) + \delta(f + (f_c + f_m)) - \frac{\beta A_c}{4} \delta(f - (f_c - f_m)) + \delta(f + (f_c - f_m)).$$



Dans ce cas ressemble à une modulation AM avec porteur (Bande latérale sup et inf) en opposition de phase.



# Synoptique de Génération



Modulation FM à large bande :  $\beta > 1$

$$g(t) = A \cos(2\pi f_c t) \cdot \cos(\beta \sin 2\pi f_m t) - A \sin(2\pi f_c t) \sin(\beta \sin 2\pi f_m t)$$

Analyse de Bessel :

$$g_{FM}(t) = A \sum_{n=-\infty}^{+\infty} J_n(\beta) \cos(2\pi(f_c - n f_m)t)$$

$$= A J_0(\beta) \cos(2\pi f_c t) + \dots$$

$$A J_1(\beta) \cos(2\pi(f_c - f_m)t) + A J_2(\beta) \cos(2\pi(f_c - 2f_m)t) + \dots$$

$$A J_{-1}(\beta) \cos(2\pi(f_c + f_m)t) + A J_{-2}(\beta) \cos(2\pi(f_c + 2f_m)t) + \dots$$

$$G(f) = \frac{A}{2} J_0 [\delta(f - f_c) + \delta(f + f_c)] - (f - (f_c - f_m))$$

$$+ \frac{A}{2} J_1 [\delta(f - (f_c - f_m)) + \delta(f + (f_c - f_m))]$$

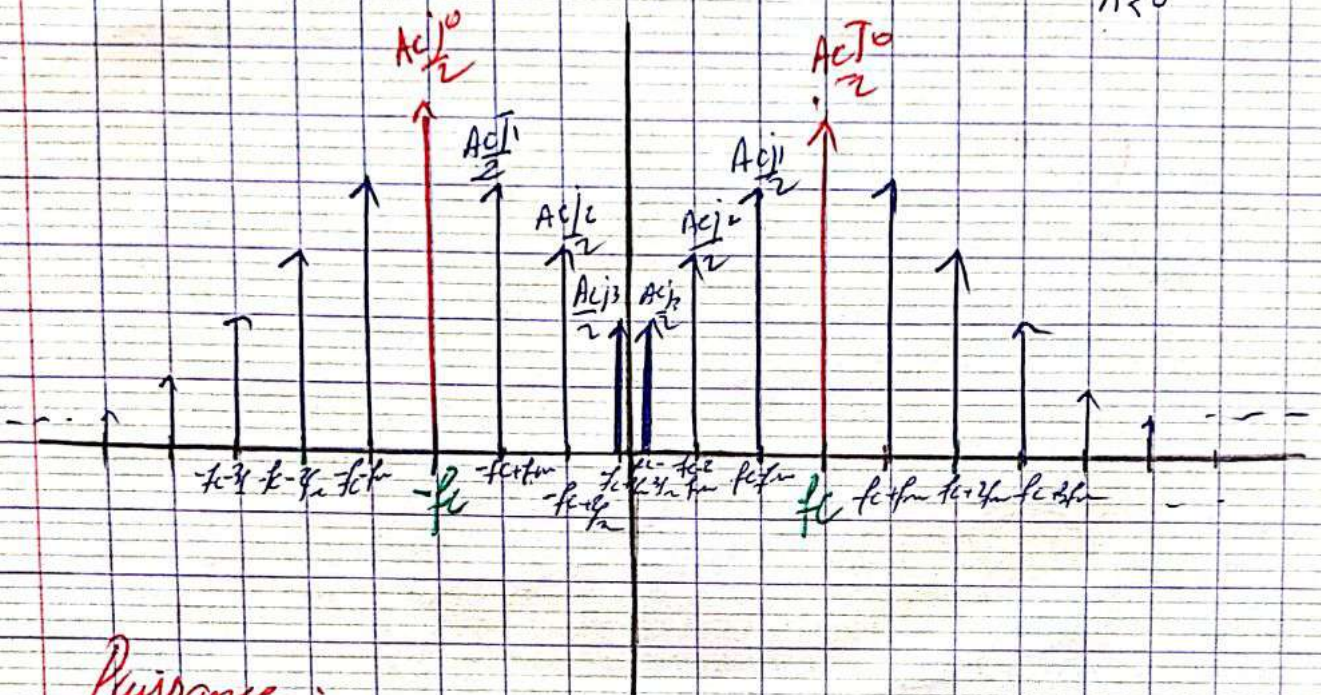
$$+ \frac{A}{2} J_2 [\delta(f - (f_c - 2f_m)) + \delta(f + (f_c - 2f_m))]$$

$$+ \frac{A}{2} J_{-1} [\delta(f - (f_c + f_m)) + \delta(f + (f_c + f_m))]$$

$$+ \frac{A}{2} J_{-2} [\delta(f - (f_c + 2f_m)) + \delta(f + (f_c + 2f_m))]$$



$n < 0$



Puissance :

$$P = \frac{V^2}{R} = \frac{V_R^2}{R}$$

$$g(t) = AC \sum_{n=-\infty}^{\infty} J_n \cos(2\pi(f_c - n f_m)t)$$

$$P = \frac{AC^2}{2R}$$